Research Article

Algebraic approaches for the design of simultaneous observers for linear systems

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Abstract: In this study, algebraic techniques are proposed to design observers capable of estimating the state of multiple linear continuous-time systems. In order to pursue this objective, first an algebraic technique is given to compute the set of all the linear inverses of the observability map of a single plant. Such a result is then used to characterise, through algebraic geometry tools, the simultaneous observability of multiple linear systems both in the forced and in the autonomous case. Such a characterisation is finally employed to design a single observer that is capable of estimating the state of multiple linear systems.

1 Introduction

In various control and identification applications, the state of a dynamical system cannot be fully measured, thus leading to the need for tools capable of estimating unmeasured variables from available measurements. Several solutions to such a problem have been proposed for both linear [[1](#page-10-0)] and non-linear [[2](#page-10-0)–[6](#page-10-0)] systems.

Although the design of a state observer for a single continuoustime system can be carried out by using classical techniques, such as the Luenberger observer [[7](#page-10-0)] and the Kalman filter [\[8\]](#page-10-0), the design of a single observer capable of estimating the state of multiple systems is much more challenging. This problem was firstly introduced in [\[9\]](#page-10-0) and can be summarised as follows: given a set of plants, design a single observer that can estimate the state of each of these plants. This problem is particularly interesting when one aims at obtaining a reliable estimate of the state of a plant affected by known perturbations that arise from sensor or actuator faults [[10\]](#page-10-0).

Several techniques have been proposed in the literature to deal with such a problem. For instance, it has been addressed in [[9](#page-10-0)], by using coprime factorisation technique, in [[11](#page-10-0)], by using evolutionary strategies, in [[12\]](#page-10-0), by using a parametrisation in terms of a stable inverse and a stable null space, and in [[13\]](#page-10-0), by using a state–space characterisation. Furthermore, in [[14,](#page-10-0) [15\]](#page-10-0), common functional observers have been proposed for two linear systems with unknown inputs. Such results have been extended in [[16\]](#page-10-0) to deal with three systems with unknown inputs. On the other hand, switching observers for switched systems have been given in [[17–](#page-10-0) [20\]](#page-10-0).

The main objective of this paper is to provide algebraic tools to design a simultaneous observer for multiple linear systems. In order to pursue this objective, in Section [3,](#page-1-0) the set of all the linear embeddings of a continuous-time system is characterised and is used to determine a parametrisation of the set of all the linear inverses of its observability map. Such a parametrisation is then used in Section [4](#page-3-0) to provide necessary and sufficient conditions for the existence of a simultaneous inverse, which holds for almost all inputs, of the observability maps of a set of linear systems. Such a result is specialised in Section [5](#page-5-0) to the case of autonomous systems, for which stronger results are obtained. In Section [6,](#page-6-0) the theoretical results established in Sections [4](#page-3-0) and [5](#page-5-0) to guarantee simultaneous observability of a set of systems are used to design a state observer capable of estimating the state of multiple plants. Several examples are given all throughout the paper to illustrate the theoretical results.

The main difference between the tools given in this paper and the ones given in $[9, 11-13]$ $[9, 11-13]$ $[9, 11-13]$ is that the former, by employing techniques borrowed from algebraic geometry, provides an exact certificate for the simultaneous observability (or lack thereof) of a set of linear continuous-time systems. Namely, given a set of systems, the proposed method allows one to determine a closedform expression for the simultaneous inverse of the observability map of such systems, if any, which can be directly used to design a simultaneous observer.

2 Notation and preliminaries

In this section, some tools of algebraic geometry are reviewed following the exposition in [\[21](#page-10-0), [22\]](#page-10-0).

Let $\mathbb{Z}, \mathbb{Z}_{\geqslant 0}, \mathbb{R}$, and $\mathbb{R}_{\geqslant 0}$ denote the sets of integer, natural, real, and non-negative real numbers, respectively. For any integer **z Notation and preliminaries**

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following the exposition in [21, 22].

Let ℤ, ℤ_{≽ 0}, ℝ, and ℝ_{≽ 0} denote the sets of integer, natural, real,

and non-ne $:= {a \in \mathbb{Z} : a \geq z}$. Symbols I_m and $\mathbf{0}_{m_1,m_2}$ denote the *m*-dimensional identity matrix and the zero matrix of dimensions Let e_i^m be the *i*th column of I_m , $i \in \{1, ..., m\}$. For any matrix $A \in \mathbb{R}^{m_1 \times m_2}$, symbol $\mathcal{E}_r(A)$ denotes the reduced row echelon form of *A*, which can be computed by using the Gauss–Jordan algorithm [[23\]](#page-10-0).

A multi-index is a vector $\mathbf{a} = [\mathbf{a}_1 \dots \mathbf{a}_n]^T \in \mathbb{Z}_{\geq 0}^n$, with *n n* Example 1. Let e_i be the ran column of I_m , t
 $A \in \mathbb{R}^{m_1 \times m_2}$, symbol $\mathcal{E}_r(A)$ denotes to f *A*, which can be computed by usin

[23].
 A multi-index is a vector $\mathbf{a} =$
 $n \in \mathbb{Z}_{\ge 1}$. Lettin \bar{x}^{α} denotes the monomial $x_1^{a_1} \ldots x_n^{a_n}$. A polynomial is a finite, linear combination of monomials; a rational function is a ratio of polynomials, with the denominator being different from the zero polynomial. The ring of all polynomials in x with coefficients in the field K is denoted $\mathbb{K}[x_1, ..., x_n]$ (briefly, $\mathbb{K}[x]$), whereas the field of all rational functions in *x* is denoted $\mathbb{K}(x_1, \ldots, x_n)$ (briefly, $\mathbb{K}(x)$). On the other hand, the set of all *m*-dimensional vectors (respectively, of all matrices of dimensions $m_1 \times m_2$) whose entries are polynomials in $\mathbb{K}[x]$ and rational functions in $\mathbb{K}(x)$ are denoted $\mathbb{K}^m[x]$ and $\mathbb{K}^m(x)$ (respectively, $\mathbb{K}^{m_1 \times m_2}[x]$ and $\mathbb{K}^{m_1 \times m_2}(x)$), respectively. matrices of dimensions $m_1 \times m_2$) whose
 $\mathbb{K}[x]$ and rational functions in $\mathbb{K}(x)$ are

(respectively, $\mathbb{K}^{m_1 \times m_2}[x]$ and $\mathbb{K}^{m_1 \times m_2}(x)$),

A subset \mathcal{F} of $\mathbb{K}[x]$ is an ideal of $\mathbb{K}[x]$

• 0 *f*(**r**espectively, $\mathbb{K}^{m_1 \times m_2}[x]$ and $\mathbb{K}^{m_1 \times m_2}(x)$)
 A subset *J* of $\mathbb{K}[x]$ is an ideal of $\mathbb{K}[$

• 0 ∈ *J*, where 0 is the zero polynomia

• if *f*, *g* ∈ *J*, then *f* + *g* ∈ *J*;

• if *f* ∈

A subset $\mathcal F$ of $\mathbb K[x]$ is an ideal of $\mathbb K[x]$ if

-
-
-

Given polynomials $p_1, ..., p_\ell \in \mathbb{K}[x]$, the set

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$$
\mathbb{V}(p_1, ..., p_\ell) := \{ \mathbf{x} \in \mathbb{K}^n : p_i(\mathbf{x}) = 0, \, i = 1, ..., \ell \}
$$

is the variety generated by $p_1, ..., p_k$

$$
(..., p_{\ell}) := \{ \mathbf{x} \in \mathbb{K}^{n} : p_{i}(\mathbf{x}) = 0, i = 1, \ldots \}
$$

generated by p_{1}, \ldots, p_{ℓ} , and the set

$$
\langle p_{1}, \ldots, p_{\ell} \rangle := \{ \sum_{i=1}^{\ell} q_{i} p_{i}, q_{i} \in \mathbb{K}[\mathbf{x}] \}
$$

is an ideal (which is referred to as the ideal generated by p_1, \ldots, p_k , ..., *p*_ℓ,

1 ideal).

(Ø) and and the set $\{p_1, ..., p_\ell\}$ is referred to as a basis of such an ideal). The sets $\{0\}$ and $\mathbb{K}[x]$ are ideals of $\mathbb{K}[x]$ and are denoted $\langle \emptyset \rangle$ and $\mathbb{K}[x]$, $\mathcal{I} \neq \langle \emptyset \rangle$, is finitely generated, i.e. there is $\{p_1, ..., p_\ell\} \subset \mathbb{K}[x]$ such that $\mathcal{I} = \langle p_1, ..., p_\ell \rangle$. The concepts of ideal and variety are strongly related. Indeed, for any ideal $\mathcal F$ in $\mathbb{K}[x]$, the variety of \mathcal{I} is

$$
\mathbb{V}(\mathcal{F}) := \{ x \in \mathbb{K}^n : p(x) = 0, \forall p \in \mathcal{F} \},
$$

and the identity

$$
\mathbb{V}(\mathcal{F})=\mathbb{V}(p_1,...,p_\ell)
$$

holds for any basis $\{p_1, \ldots, p_\ell\}$ of \mathcal{I} . Similarly, for any subset and the identity

holds for any basis {*I*
 $\mathcal{R} \subset \mathbb{K}^n$, the set
 $\mathbb{I}(\mathcal{R}) := \{I\}$ $n₁$, the set

$$
\mathbb{I}(\mathcal{R}) := \{ p \in \mathbb{K}[x] : p(x) = 0, \forall x \in \mathcal{R} \}
$$

is an ideal of $K[x]$ even if $\mathcal R$ is not a variety (such an ideal is referred to as the ideal of \mathcal{R}). For any ideal \mathcal{I} of $\mathbb{K}[\mathbf{x}]$, one has that $\mathcal{I} \subset \mathbb{I}(\mathbb{V}(\mathcal{I}))$, but the converse inclusion need not hold, unless \mathbb{K} is $\mathbb{I}(\mathcal{R}) := \{p \in \mathbb{K}[\mathbf{x}] : p(\mathbf{x}) = 0, \forall \mathbf{x} \in \mathcal{R}\}\$
is an ideal of $\mathbb{K}[\mathbf{x}]$ even if \mathcal{R} is not a variety (such an ideal is
referred to as the ideal of \mathcal{R}). For any ideal $\mathcal{I} \circ f \mathbb{K}[\mathbf{x}]$, one n^n , one has that $\mathscr{R} \subset \mathbb{V}(\mathbb{I}(\mathscr{R}))$. In particular, the set $\mathscr{LR} := \mathbb{V}(\mathbb{I}(\mathscr{R}))$ is the smallest variety in K^n that contains \mathcal{R} , and is called the Zariski basis for each sub-module in $K^m[\mathbf{x}]$, closure of \mathcal{R} . A property holds for 'almost all' $x \in \mathbb{K}^n$ (or, been equivalently, it is 'generic' in $Kⁿ$) if the Zariski closure of the set

A monomial order $>$ on $\mathbb{K}[x]$ is a total, well ordering relation on the set of monomials $x^a \in \mathbb{K}[x]$. The lexicographic order (briefly, the Lex order), denoted \succ_L , is a monomial order and is defined as follows: $x^a >_L x^b$ if in the vector difference $a - b$ the cor first non-zero entry is positive. Hence, let any monomial order ≻ be fixed. For any $p \in \mathbb{K}[x]$, the leading term of p, denoted $LT(p)$, is the greatest term $c \mathbf{x}^{\mathfrak{a}}$ appearing in *p*. A polynomial $r \in \mathbb{K}[\mathbf{x}]$ is reduced with respect to $\{p_1, ..., p_\ell\}$ if either $r = 0$ or no monomial enoted ≻_L, is a monomial order and is
 $\mathbf{x}^{\mathbf{b}}$ if in the vector difference $\mathbf{a} - \mathbf{b}$ the

tive. Hence, let any monomial order ≻
 j, the leading term of *p*, denoted LT(*p*),

pearing in *p*. A polynomial of *r* is divisible by any $LT(p_i)$, $i = 1, ..., \ell$. A finite set $\{g_1, ..., g_{\tau}\}\)$ of $\mathbb{K}[x]$ is a *Gröbner basis* of an ideal \mathcal{I} of $\mathbb{K}[x]$ if

$$
\langle LT(g_1), ..., LT(g_\tau) \rangle = \langle \{c\,x^{\mathfrak{a}} : \exists f \in \mathcal{F} : LT(f) = c\,x^{\mathfrak{a}} \} \rangle.
$$

A Gröbner basis $\{g_1, \ldots, g_\tau\}$ is reduced if g_i is reduced with respect to $\{g_1, ..., g_{i-1}, g_{i+1}, ..., g_{\tau}\}\$ and $LT(g_i) = x^a$, for some $a \in \mathbb{Z}_{\geq 0}^n$, (x) if
 $L T(f) = c x^a$ }). derivat

s reduced with respect

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s for some $\mathbf{a} \in \mathbb{Z}_{\geq 0}^n$, $\psi_{e,N}(t)$

ique reduced Gröbner

consider $i = 1, \ldots, \tau$. Each ideal $\mathcal F$ of $\mathbb K[x]$ has a unique reduced Gröbner basis. A Gröbner basis $\{g_1, ..., g_r\}$ is reduced if g_i is reduced with respect
to $\{g_1, ..., g_{i-1}, g_{i+1}, ..., g_r\}$ and $LT(g_i) = x^a$, for some $\mathbf{a} \in \mathbb{Z}_{\geq 0}^n$,
 $i = 1, ..., \tau$. Each ideal \mathcal{F} of $\mathbb{K}[x]$ has a unique reduced to $\{g_1, ..., g_{i-1}, g_{i+1}, ..., g_r\}$ and $LT(g_i) = x^a$, for some **a**
 $i = 1, ..., \tau$. Each ideal $\mathcal F$ of $\mathbb{K}[x]$ has a unique reduced (basis.

Given two ideals $\mathcal F$ and $\mathcal F$ of $\mathbb{K}[x]$, the intersection of
 $\mathcal F$, denoted

Given two ideals $\mathcal F$ and $\mathcal J$ of $\mathbb K[x]$, the intersection of $\mathcal J$ and

$$
\mathcal{F} \cap \mathcal{J} = (t \mathcal{J} + (1 - t) \mathcal{J}) \cap \mathbb{K}[x],
$$

where *t* is a single auxiliary variable.

The notion of ideal can be generalised to deal with polynomial vectors in $K^m[x]$. Namely, given polynomial vectors $p_1, ..., p_\ell \in \mathbb{K}^m[x]$, the sub-module of $\mathbb{K}^m[x]$ generated by $\boldsymbol{p}_1, \ldots, \boldsymbol{p}_{\ell}$ is

$$
\langle \boldsymbol{p}_1,\ldots,\boldsymbol{p}_{\ell}\rangle=\left\{\sum_{i=1}^{\ell}q_i\,\boldsymbol{p}_i,\,q_i\in\mathbb{K}[\boldsymbol{x}]\right\}.
$$

Similarly, given a polynomial matrix $R = [r_1 \dots r_{m_2}] \in \mathbb{K}^{m_1 \times m_2}[x]$, define the image of *R* as a submodule of $\mathbb{K}^{m_1}[x]$,

$$
\mathrm{Img}(\boldsymbol{R}) := \langle \boldsymbol{r}_1, \ldots, \boldsymbol{r}_{m_2} \rangle.
$$

On the other hand, for any polynomial matrix $\mathbf{R} \in \mathbb{K}^{m_1 \times m_2}[x]$, the syzygy of **R** is the sub-module of $\mathbb{K}^{m_2}[x]$ consisting of all polynomial vectors $p \in \mathbb{K}^{m_2}$ (*x*) $:= \langle r_1, ..., r_{m_2} \rangle$.

polynomial matrix $R \in \mathbb{K}^{m_1 \times m_2}[x]$, the

module of $\mathbb{K}^{m_2}[x]$ consisting of all

[*x*] such that $R p = \mathbf{0}_{m_1, 1}$, where $\mathbf{0}_{m_1, 1}$ is

onomial order can be extended to the

o the zero polynomial vector.

VR. — a). For ± 1 C interpretation μ and μ is the Hilbert basis theorem, each ideal μ is the Hilbert Basis of the Hilbert Basis theorem, each interpretation μ is the Hilbert Basis of the Hilbert Basis of Also, the concept of monomial order can be extended to the deal with polynomial vectors in $K^m[x]$. Indeed, a monomial in $\mathbb{K}^m[x]$ is a product of the form $x^a e_i^m$, so that each p in $\mathbb{K}^m[x]$ can be written as a finite linear combination, with coefficients in K , of monomials in $\mathbb{K}^m[x]$. Hence, a monomial order > on $\mathbb{K}^m[x]$ is a total, well ordering relation on the set of monomials in $K^m[x]$. For instance, the POT extension of the Lex order is a monomial order on $\mathbb{K}^m[x]$ and is defined as follows: $x^a e_i^m >_L x^b e_j^m$ if $i < j$ or $i = j$ is and $x^a >_L x^b$. Hence, let any monomial order on K^{*m*}[x] be fixed.

The leading term of $p \in \mathbb{K}^m[x]$, denoted LT(p), is the greatest term
 $c x^a e_i^m$ appearing in p . Given a sub-module *M* of K^{*m*}[x], let The leading term of $p \in \mathbb{K}^m[x]$, denoted $LT(p)$, is the greatest term $c x^a e_i^m$ appearing in *p*. Given a sub-module *M* of $\mathbb{K}^m[x]$, let $LT(M)$ be the sub-module generated by the leading terms of all *m* [*x*], a finite set $\{g_1, ..., g_\tau\}$ of $\mathbb{K}^m[x]$ is a Gröbner basis of M if

$$
\langle LT(\mathcal{M})\rangle = \langle LT(\boldsymbol{g}_1), ..., LT(\boldsymbol{g}_{\tau})\rangle.
$$

Reduced Gröbner bases of sub-modules can be defined as for ideals, and it can be shown that there is a unique reduced Gröbner basis for each sub-module in $K^m[x]$, once a monomial order has been fixed.

where such a property 'does not hold' does not coincide with \mathbb{K}^n . Sumput-output embeddings or single-input 3 Input-output embeddings of single-input **single-output (SISO) continuous-time linear systems**

Consider the following single-input single-output (briefly, SISO) continuous-time linear system

$$
\dot{\xi}(t) = E \, \xi(t) + F \, \nu(t), \quad \psi(t) = G \, \xi(t) + H \, \nu(t), \tag{1}
$$

where $\xi(t) \in \mathbb{R}^n$ is the state vector, $\nu(t) \in \mathbb{R}$ is the scalar input, and $\psi(t) \in \mathbb{R}$ is the scalar output at time $t \in \mathbb{R}_{\geq 0}$; $\nu(t)$ is assumed to be is the state vector, *v*(*t*) = *G* ξ(*t*) + *H v*(*t*), SISO)

inear system
 $\xi(t) + F v(t)$, *v*(*t*) = *G* ξ(*t*) + *H v*(*t*), (1)

is the state vector, *v*(*t*) $\in \mathbb{R}$ is the scalar input, and

calar output at ti *ξ*(*t*) = *E ξ*(*t*) + *F ν*(*t*), *ψ*(*t*) = *G ξ*(*t*) + *H ν*(*t*), (1)
where *ξ*(*t*) ∈ ℝ^{*n*} is the state vector, *ν*(*t*) ∈ ℝ is the scalar input, and
 $ψ(t) ∈ ℝ$ is the scalar output at time *t* ∈ ℝ_{≥0}; *ν*(differentiable a sufficiently high number of times. Let $\nu^{(i)}(t) := d^i \nu(t) / dt$ and $\psi^{(i)}(t) := d^i \psi(t) / dt$ denote the *i*th timewhere $\xi(t) \in \mathbb{R}^n$ is the state vector, $\nu(t) \in \mathbb{R}$ is the scalar input, and $\psi(t) \in \mathbb{R}$ is the scalar output at time $t \in \mathbb{R}_{\geq 0}$; $\nu(t)$ is assumed to be differentiable a sufficiently high number of times any $N \in \mathbb{Z}_{\geq 0}$, let $\nu_{e,N}(t) = \left[\nu^{(0)}(t) \dots \nu^{(N)}(t) \right]^\top$ be a vector having as entries the input and its time-derivatives and let $\psi_{e,N}(t) = [\psi^{(0)}(t) \dots \psi^{(N)}(t)]^{\top}$ be a vector having as entries the corresponding output response and its time-derivatives. Hence, consider the following definition (see [\[24](#page-10-0), [25\]](#page-10-0)).

Definition 1: A polynomial $p \in \mathbb{R}[\psi_{e,N}, \nu_{e,N}]$ is an embedding of system (1) if

 $p(\psi_{e,N}(t), \nu_{e,N}(t)) = 0, \quad \forall t \in \mathbb{R}_{\geq 0}.$

An embedding is linear if $deg(p) = 1$.

In order to characterise the set of all embeddings of system (1), which is an ideal of $\mathbb{R}[\psi_{e,N}, \nu_{e,N}]$ by of [\[24](#page-10-0), [26\]](#page-10-0), define, for all $N \in \mathbb{Z}_{\geq 0}$, the matrices

$$
M_N := \begin{bmatrix} H & 0 & \dots & 0 \\ GF & H & \dots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ GE^{N-1}F & GE^{N-2}F & \dots & H \end{bmatrix}, \qquad (2a)
$$

$$
O_N := \begin{bmatrix} G \\ GE \\ \vdots \\ GE^N \end{bmatrix} . \qquad (2b)
$$

Thus, consider the following lemma.

Lemma 1: For any $N \in \mathbb{Z}_{\geq 0}$, letting $\xi(t)$ be the state response of system ([1](#page-1-0)), one has

$$
\psi_{e,N}(t) = O_N \xi(t) + M_N \nu_{e,N}(t), \quad \forall t \in \mathbb{R}_{\geq 0}.
$$
 (3)

Proof: The proof follows directly from the analysis carried out in Section 5.1.2 of [\[27](#page-10-0)] and from Proposition 1 of [\[28](#page-10-0)]. \Box

Given $N \in \mathbb{Z}_{\geq 0}$, $N \geq n - 1$, define the observability map of order *N* of system [\(7\)](#page-3-0),

$$
\Psi_N(\xi,\nu_{e,N}) := O_N \xi + M_N \nu_{e,N},
$$

which, by Lemma 1, relates the current state of system [\(7\)](#page-3-0) and the time-derivatives of its input ν up to order *N*, with the timederivatives of the output ψ up to order *N*. Hence, in order to determine the set of all embeddings of system [\(7\)](#page-3-0), define the matrix

$$
\mathbf{Q}_N := \begin{bmatrix} \mathbf{O}_N & -\mathbf{I}_{N+1} & \mathbf{M}_N \end{bmatrix} \in \mathbb{R}^{(N+1) \times (2N+n+2)},
$$

which, by (3), is such that

$$
\mathbf{Q}_N\begin{bmatrix} \boldsymbol{\xi}(t) \\ \boldsymbol{\psi}_{e,N}(t) \\ \boldsymbol{\nu}_{e,N}(t) \end{bmatrix} = \mathbf{0}_{N+1,1}, \quad \forall t \in \mathbb{R}_{\geq 0}.
$$
 (4)

Thus, let $\mathbf{Z}_N := \mathcal{E}_r(\mathbf{Q}_N)$ be the reduced row echelon form of \mathbf{Q}_N and partition such a matrix as follows:

n columns
$$
2N + 2
$$
 columns $Z_N = \begin{bmatrix} Z_{N,1,1} \\ \mathbf{0}_{\nu,n} \end{bmatrix}$ $Z_{N,2,2}$

i.e., let $Z_{N,2,2} \in \mathbb{R}^{\nu \times (2N+2)}$ be the matrix obtained by retaining only the last $2N + 2$ columns of Z_N and the rows that contain a pivot. Thus, consider the following theorem.

Theorem 1: Let \mathbf{Z}_N be partitioned as above. The set of all linear embeddings of system [\(1\)](#page-1-0) in $\mathbb{R}[\psi_{e,N}, \nu_{e,N}]$ is

ngs of system (1) in
$$
\mathbb{R}[\psi_{e,N}, \nu_{e,N}]
$$
 is
\n
$$
\mathcal{L} = \left\{ \sum_{j=1}^{\mu} c_j q_j, \forall c_j \in \mathbb{R} \right\},
$$
\n
$$
\dots, q_{\mu}
$$
 are given by\n
$$
[q_1 \quad \dots \quad q_{\mu}]^{\top} = \mathbf{Z}_{N,2} \left\{ \psi_{e,N} \right\}
$$

where q_1, \ldots, q_u are given by

$$
\begin{bmatrix} q_1 & \dots & q_\mu \end{bmatrix}^\top = \mathbf{Z}_{N,2,2} \begin{bmatrix} \mathbf{\Psi}_{e,N} \\ \mathbf{\nu}_{e,N} \end{bmatrix}.
$$

Furthermore, $\mathcal{J}_N := \langle q_1, ..., q_\mu \rangle$ is the ideal of all embeddings in $\mathbb{R}[\psi_{e,N}, \nu_{e,N}]$ of system ([1](#page-1-0)).

Proof: By [\[23](#page-10-0)], there exists a matrix Λ_N such that $Z_N = \Lambda_N Q_N$, and, therefore, it results that,

$$
\mathbf{Z}_N\begin{bmatrix} \boldsymbol{\xi}(t) \\ \boldsymbol{\psi}_{e,N}(t) \\ \boldsymbol{\nu}_{e,N}(t) \end{bmatrix} = \mathbf{0}_{\nu,1}, \quad \forall t \in \mathbb{R}_{\geq 0}.
$$

In particular, this implies that the polynomials q_1, \ldots, q_l $, ..., q_{\mu}$ are linear,
no embedding of
because the steps embeddings of system [\(1\)](#page-1-0). Note that there is no embedding of system [\(1\)](#page-1-0) in $\mathbb{R}[\psi_{e,N}, \nu_{e,N}]$ that is not in \mathcal{J}_N , because the steps carried out to compute $Z_N = \mathcal{E}_r(Q_N)$ are exactly those to compute the reduced Gröbner basis of the ideal generated by the relations *ξ*¹ ≻^L …*ξⁿ* ≻^L *ψ* (0) ≻^L … ≻^L *ψ* (*N*) ≻^L *ν* (0) ≻^L … ≻^L *ν* (*N*) [\[21](#page-10-0)] via the Buchberger's algorithm [\[29](#page-10-0), [30](#page-10-0)]. Hence, the statement follows directly from Theorem 2 of [[24\]](#page-10-0). □

Remark 1: If pair (E, G) is observable, then matrix $Z_{N,2,2}$ has exactly $N - n + 1$ rows due to the fact that the matrix O_N has rank *n*, and hence $\mathbf{Z}_{N,1,1} = \mathbf{I}_n$.

By combining the results established in Lemma 1 and in Theorem 1, the set of all linear inverses of the observability map of system [\(1\)](#page-1-0) is characterised by the following corollary.

Corollary 1: Assume that pair (*E*, *G*) is observable. For any $N \in \mathbb{Z}_{\geq 0}$, $N \geq n$, the set of all linear inverses of the observability map $\Psi_N(\cdot, \cdot)$ of system ([7](#page-3-0)), i.e. the set of all linear functions $\Xi_N(\cdot,\cdot)$ such that

$$
\Xi_N(\Psi_N(\xi,\nu_{e,N}),\nu_{e,N})=\xi,\quad \forall \xi\in\mathbb{R}^n, \forall \nu_{e,N}\in\mathbb{R}^{N+1},
$$

is parametrised by

$$
\mathbf{\Xi}_{N}(\boldsymbol{\psi}_{e,N},\boldsymbol{\nu}_{e,N}) = \boldsymbol{O}_{n-1}^{-1}(\boldsymbol{\psi}_{e,n-1} - \boldsymbol{M}_{n-1}\boldsymbol{\nu}_{e,n-1}) + \sum_{j=1}^{N-n+1} c_j q_j(\boldsymbol{\psi}_{e,N},\boldsymbol{\nu}_{e,N}),
$$
\n(5)

where the polynomials q_1, \ldots, q_{N-n+1} are defined as in Theorem 1 and $c_1, ..., c_{N-n+1}$ are arbitrary vectors in ℝⁿ. *n*

[8) Z_0 (**8**) \mathbf{v}_0 (**8**) \mathbf{v}_1 (**8**) \mathbf{v}_2 (**8**) \mathbf{v}_3 (**8**) \mathbf{v}_4 (**8**) and the problems in (**a** in equivale *s* (**a** in equivale in the Lex order with \mathbf{v}_2 (**b** in equivale **a** in equivale *Proof:* By Remark 1, if pair (E, G) is observable, then $\nu = N - n + 1$. In view of Lemma 1, the set of all linear formulas relating $\xi(t)$ with $\psi_{e,N}(t)$ and $\nu_{e,N}(t)$ corresponds to the set of all solutions in *ξ* to the system of equations given in (3). Since pair (*E*, *G*) is observable, one of the formulas that relates *ξ*(*t*) with $\psi_{e,N}(t)$ and $\nu_{e,N}(t)$ is $\xi(t) = O_{n-1}^{-1}(\psi_{e,n-1}(t) - M_{n-1}\nu_{e,n-1}(t))$, which corresponds to (5) with $c_1 = \cdots = c_\mu = 0$. Thus, since (4) is a system of linear equations that hold for all times $t \in \mathbb{R}_{\geq 0}$, the relating $\xi(t)$ with $\psi_{e,N}(t)$ and $\nu_{e,N}(t)$ corresponds to the set of all
solutions in ξ to the system of equations given in (3). Since pair
 (\mathbf{E}, \mathbf{G}) is observable, one of the formulas that relates $\xi(t)$ with formula given in (5) follows by the facts that, by Theorem 1, \mathcal{L} is the set of all the linear embeddings of system [\(1\)](#page-1-0) and that the set of all the solutions to (4) is given by the sum of a particular solution to (4) and the set of all the solutions of the corresponding homogeneous equation [\[23](#page-10-0)]. □

The expression in (5) can be used also to find non-linear formulas relating to the current state $\xi(t)$ of system ([1](#page-1-0)) with $\psi_{e,N}(t)$ and $\nu_{e,N}(t)$. Indeed, by a trivial extension of Corollary 1, non-linear formulas relating the state and the time-derivative of the input and of the output are given by

$$
\Sigma_N(\psi_{e,N}, \nu_{e,N}) = O_{n-1}^{-1}(\psi_{e,n-1} - M_{n-1}\nu_{e,n-1}) + \sum_{j=1}^{N-n+1} c_j(\psi_{e,N}, \nu_{e,N}) q_j(\psi_{e,N}, \nu_{e,N}),
$$
(6)

where $c_j(\psi_{e,N}, \nu_{e,N})$ are arbitrary vector functions of $\psi_{e,N}$ and $\nu_{e,N}$. The expression given in (6) is exploited in the following Section [4](#page-3-0) to characterise the simultaneous observability of a set of linear systems.

The following example illustrates the application of Corollary 1.

Example 1: Consider system [\(1](#page-1-0)) with

$$
E = \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix}, \quad F = \begin{bmatrix} 0 \\ 1 \end{bmatrix}, \quad G = \begin{bmatrix} 1 & 0 \end{bmatrix}, \quad D = 1.
$$

Example 1: Consider system (1) with
 $E = \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix}$, $F = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$, $G = \begin{bmatrix} 1 & 0 \end{bmatrix}$, $D = 1$.

Letting $N = 3$ and by using the result established in Corollary 1, it

turns out that the set of all turns out that the set of all linear inverses of the observability map $E = \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix}$, $F = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$, $G = \begin{bmatrix} 1 \end{bmatrix}$
Letting $N = 3$ and by using the result establis
turns out that the set of all linear inverses of t
 $\Psi_3(\cdot, \cdot)$ of system (7) can be parametrised as

$$
\mathbf{\Xi}_{3}(\boldsymbol{\psi}_{e,3},\boldsymbol{\nu}_{e,3}) = \begin{bmatrix} \psi^{(0)} - \nu^{(0)} \\ \psi^{(1)} - \nu^{(1)} \end{bmatrix} + \mathbf{c}_{1}(\psi^{(1)} + \psi^{(3)} - 2\nu^{(1)} - \nu^{(3)}) + \mathbf{c}_{2}(\psi^{(0)} + \psi^{(2)} - 2\nu^{(0)} - \nu^{(2)}),
$$

2

4 Simultaneous inverse of the observability map for multiple SISO continuous-time linear systems

Consider the SISO continuous-time linear system:

$$
\dot{x}(t) = A x(t) + B u(t), \quad y(t) = C x(t) + D u(t), \tag{7}
$$

where $x(t) \in \mathbb{R}^n$, $u(t) \in \mathbb{R}$, and $y(t) \in \mathbb{R}$ denote the state vector, the input, and the output, respectively, at time $t \in \mathbb{R}_{\geq 0}$; $u(t)$ is assumed SO continuous-time linear systems

SO continuous-time linear system:
 $u(x(t) + Bu(t), y(t) = Cx(t) + Du(t),$ (7)
 $u(t) \in \mathbb{R}$, and $y(t) \in \mathbb{R}$ denote the state vector, the

striput, respectively, at time $t \in \mathbb{R}_{\geq 0}$; $u(t)$ is Consider the SISO continuous-time linear system:
 $\dot{x}(t) = A x(t) + B u(t),$ y(*t*) = *C* x(*t*) + *D u*(*t*), (7)

where $x(t) ∈ ℝ^n$, *u*(*t*) ∈ ℝ, and y(*t*) ∈ℝ denote the state vector, the

input, and the output, respectively, to be differentiable a sufficiently high number of times. Assume tuple (A, B, C, D) belongs to the following finite set:

$$
\mathcal{S} := \{ (A_1, B_1, C_1, D_1), ..., (A_s, B_s, C_s, D_s) \},
$$

that *A*, *B*, *C*, and *D* are not known, but that it is known that the
tuple (*A*, *B*, *C*, *D*) belongs to the following finite set:
 $\mathcal{S} := \{ (A_1, B_1, C_1, D_1), ..., (A_s, B_s, C_s, D_s) \}$,
with $s \in \mathbb{Z}_{\ge 2}$.
The main objec The main objective of this section is, given the set S , to compute a function, if any, that relates the current state $x(t)$ of system (7) with the time-derivatives of *u*(*t*) and *y*(*t*) up to order *N* $(u_{e,N} = [u^{(0)} \dots u^{(N)}]^T$, $u^{(i)}(t) = d^i u(t) / dt^i$, $i = 0, ..., N$ and $y_{e,N} = [y^{(0)} \dots y^{(N)}]^\top$, $y^{(i)}(t) = d^i y(t) / dt^i$, $i = 0, ..., N$, respectively), for some $N \in \mathbb{Z}_{\geq 0}$.

By Lemma 1, letting $M_{k,N}$ and $O_{k,N}$ be the matrices obtained by substituting, in the expressions given in (2) for M_N and O_N , E with A_k , *F* with B_k , *G* with C_k , and *H* with D_k , $k = 1, ..., s$, it results that there exists $k \in \{1, ..., s\}$ such that

$$
\mathbf{y}_{e,N}(t) = \boldsymbol{O}_{k,N}\,\boldsymbol{x}(t) + \boldsymbol{M}_{k,N}\,\boldsymbol{u}_{e,N}(t)\,.
$$

Thus, define the observability map of order *N* of the *k*th system,

$$
\Psi_{k,N}(x,u_{e,N}) := O_{k,N} x + M_{k,N} u_{e,N}, \qquad (8)
$$

which relates the current state of system [\(1\)](#page-1-0) and the timederivatives of $u(t)$ and $y(t)$ up to order *N*, provided that $A = A_k$, $\mathbf{B} = \mathbf{B}_k$, $\mathbf{C} = \mathbf{C}_k$, and $D = D_k$. Therefore, the main objective of this section is to compute, if any, a function $\Phi_N(\cdot, \cdot)$ such that

$$
\mathbf{\Phi}_N(\mathbf{\Psi}_{k,N}(x,u_{e,N}),u_{e,N})=x,
$$

 $\mathbf{B} = \mathbf{B}_k$, $\mathbf{C} = \mathbf{C}_k$, and $D = D_k$. T
section is to compute, if any, a f
 $\mathbf{\Phi}_N(\mathbf{\Psi}_{k,N}(x))$, for 'almost all' $\mathbf{x} \in \mathbb{R}^n$, $\mathbf{u}_{e,N} \in$
such a function exists, then it is
observability maps $\mathbf{\Psi}_{k$ $, u_{e,N} \in \mathbb{R}^{N+1}$, and for all $k \in \{1, ..., s\}$. If such a function exists, then it is called *simultaneous inverse of the observability maps* $\Psi_{k,N}(\cdot, \cdot)$ *, k* = 1, …, *s*.

Assumption 1: Pair (A_k, C_k) is observable, $k = 1, ..., s$.

Clearly, if Assumption 1 does not hold, then there does not exist a simultaneous inverse of the observability maps $\Psi_{k,N}(\cdot, \cdot)$, $k = 1, \ldots, s$, due to the fact that the relation given in (8) is not invertible for some $k \in \{1, ..., s\}$. In addition, as well known, 'almost all' pairs (A_k, C_k) are observable.

Under Assumption 1, let $N \in \mathbb{Z}_{\geq 0}$, $N \geq n$, and let

$$
\mathcal{G}_{k,N} := \{g_{k,N,1}, \dots, g_{k,N,\ell}\},\tag{9}
$$

y⁽⁰⁾ ≻_L ... ≻_L *y*^(*N*) ≻_L *u*⁽⁰⁾ ≻_L ... ≻_L *u*^(*N*) of the set all embeddings of

$$
\dot{\chi}_k(t) = A_k \chi_k(t) + B_k u(t),
$$

\n
$$
y(t) = C_k \chi_k(t) + D_k u(t),
$$
\n(10)

for $k = 1, ..., s$. Note that, by Remark 1, the set $\mathcal{G}_{k,N}$ is composed by exactly $\ell := N - n + 1$ linear polynomials and can be easily computed by using the method illustrated just above Theorem 1. Hence, for $k = 1, \ldots, s$, let

where
$$
c_1
$$
 and c_2 are arbitrary vectors in \mathbb{R}^2 .
\n4 Simultaneous inverse of the observability map\n
$$
\begin{aligned}\nJ_{k,N} &:= I_n \otimes [g_{k,N,1} \cdots g_{k,N,\ell}] \\
&= \begin{bmatrix}\ng_{k,N,1} & \cdots & g_{k,N,\ell} \\
\vdots & \vdots & \ddots & \vdots \\
0 & 0 & 0 & \cdots & g_{k,N,1} \cdots & g_{k,N,\ell}\n\end{bmatrix}, \\
&= \begin{bmatrix}\ng_{k,N,1} & \cdots & g_{k,N,\ell} \\
\vdots & \vdots & \ddots & \vdots \\
0 & 0 & 0 & \cdots & g_{k,N,1} \cdots & g_{k,N,\ell}\n\end{bmatrix}.\n\end{aligned}
$$

and define the following vector (being linear in $y_{e,n-1}$ and $u_{e,n-1}$)

$$
\theta_k := O_{k,n-1}^{-1} (y_{e,n-1} - M_{k,n-1} u_{e,n-1}).
$$
\n(11)

Define the matrices

$$
W_N := \begin{bmatrix} J_{1,N} & -J_{2,N} & \dots & 0_{n,n\ell} \\ \vdots & \vdots & \ddots & \vdots \\ J_{1,N} & 0_{n,n\ell} & \dots & -J_{s,N} \end{bmatrix},
$$
(12a)

$$
\mathbf{\Theta} := \begin{bmatrix} \boldsymbol{\theta}_1 - \boldsymbol{\theta}_2 \\ \vdots \\ \boldsymbol{\theta}_1 - \boldsymbol{\theta}_s \end{bmatrix},\tag{12b}
$$

and consider the following theorem.

Theorem 2: Let *S* be given and $N \in \mathbb{Z}_{\geq n}$; under Assumption 1, \overline{S} consider the following sub-module of $\mathbb{R}^{n(s-1)}[y_{e,N}, u_{e,N}]$:

$$
\mathscr{P}:=\langle [W_N \quad \Theta]\rangle.
$$

 \mathcal{L}_{2} , \mathcal{L}_{3} = 1 (*R, w*, ..., ..., *R*, *w*, ...)

be the reduced Grobner basis according to the Lex order with \mathcal{L}_{N} ... \mathcal{L}_{2} , \mathcal{L}_{3} , \mathcal{L}_{3} , \mathcal{L}_{4} , \mathcal{L}_{5} , \mathcal{L}_{6} , \mathcal{L}_{7} Let $S_N \in \mathbb{R}^{(n\ell s + 1) \times \varrho}[y_{e,N}, u_{e,N}]$ be the reduced Gröbner basis of the syzygy of $[W_N \quad \Theta]$ with respect to the POT extension of the Lex order with $y^{(0)} \succ_L ... \succ_L y^{(N)} \succ_L u^{(0)} \succ_L ... \succ_L u^{(N)}$. Thus, there exists a simultaneous inverse $\Phi_N(\cdot, \cdot)$ of the observability maps $\Psi_{k,N}(\cdot, \cdot)$, $k = 1, ..., s$, that holds for 'almost all' $u_{e,N}$ if and only Let $S_N \in \mathbb{R}^N$ $\{V_{e,N}, u_{e,N}\}$ be the reduced Grobner basis of the
syzygy of $[W_N \Theta]$ with respect to the POT extension of the Lex
order with $y^{(0)} >_L ... >_L y^{(N)} >_L u^{(0)} >_L ... >_L u^{(N)}$. Thus, there
exists a simultaneous i can be partitioned as

$$
S_N = \begin{bmatrix} S_{N,1,1} | S_{N,1,2} | S_{N,1,3} | n \, \ell \, s \text{ rows} \\ \mathbf{0}_{1,\rho} \, & \, \mathbf{0} \, S_{N,2,3} \end{bmatrix} \begin{matrix} n \, \ell \, s \text{ rows} \\ 1 \text{ row} \end{matrix}.
$$

In such a case, by letting $S_{N,1,2}$ be partitioned as

$$
S_{N,1,2} = \begin{bmatrix} S_{N,1,2,1} \\ \vdots \\ S_{N,1,2,s} \end{bmatrix}_{n \ell \text{ rows}}^{n \ell \text{ rows}},
$$

a simultaneous inverse of the observability maps $\Psi_{k,N}(\cdot, \cdot)$, $k = 1, \ldots, s$, holding for 'almost all' $u_{e,N}$, is

$$
\mathbf{\Phi}_N = \boldsymbol{\theta}_1 + \frac{1}{\omega}(\boldsymbol{J}_{1,N}\boldsymbol{S}_{N,1,2,1}).
$$
\n(13)

Proof: Since S_N is a syzygy matrix of the polynomial submodule \mathcal{P} , whose basis is $[W_N \quad \Theta]$, it holds that

In particular, by the definition of the polynomial matrices W_N and **Θ**, the following polynomial relations hold:

$$
\frac{1}{\omega} \mathbf{J}_{1,N} \mathbf{S}_{N,1,2,1} + \theta_1 = \frac{1}{\omega} \mathbf{J}_{2,N} \mathbf{S}_{N,1,2,2} + \theta_2,
$$
\n
$$
\vdots
$$
\n
$$
\frac{1}{\omega} \mathbf{J}_{1,N} \mathbf{S}_{N,1,2,1} + \theta_1 = \frac{1}{\omega} \mathbf{J}_{s,N} \mathbf{S}_{N,1,2,s} + \theta_s.
$$

Hence, letting $\Phi_N(y_{e,N}, u_{e,N})$ be defined as in [\(13](#page-3-0)), by Theorem 1, $\Phi_N(\cdot, \cdot)$ is a simultaneous inverse of the observability maps Ψ_{k} , $N(\cdot, \cdot)$, $k = 1, ..., s$. Furthermore, it holds for 'almost all' \mathbf{u}_{k} , because the set where it cannot be applied is the variety V (*ω*) ⊂ ℝ^{*N*+1}, which does not coincide with ℝ^{*N*+1} since *ω* ≠ 0. (13), by Theorem 1,
observability maps
for 'almost all' $u_{e,N}$
lied is the variety
 x^{N+1} since $\omega \neq 0$.
xists a simultaneous
maps $\Psi_{k,N}(\cdot, \cdot)$,
harang by ω , ω , ω

On the other hand, assume that there exists a simultaneous inverse $\Phi_N(\cdot, \cdot)$ of the observability maps $\Psi_{k,N}(\cdot, \cdot)$, $k = 1, \ldots, s$, that holds for 'almost all' $u_{e,N}$. By Theorem 1, since $\Phi_N(\cdot, \cdot)$ can be used for 'almost all' $u_{e,N}$, there exist $c_{k,j} \in \mathbb{R}^n$ (*y*_{*e*,*N*}, *u*_{*e*,*N*}) whose entries are rational functions with numerator in ℝ[$y_{e,N}$, $u_{e,N}$] and denominator in ℝ[$u_{e,N}$] such that

$$
\Phi_N = O_{k,n-1}^{-1} (y_{e,n-1} - M_{k,n-1} u_{e,n-1}) + \sum_{j=1}^{N-n+1} c_{k,j} g_{k,N,j},
$$

for $k = 1, \ldots, s$. By taking pairwise differences of the expression above for $k \in \{1, ..., s\}$, one obtains that

$$
\theta_1 - \theta_2 + \sum_{j_1=1}^{N-n+1} c_{1,j_1} g_{1,j_1} - \sum_{j_2=1}^{N-n+1} c_{2,j_2} g_{2,j_2} = 0,
$$

\n
$$
\vdots
$$

\n
$$
\theta_1 - \theta_3 + \sum_{j_1=1}^{N-n+1} c_{1,j_1} g_{1,j_1} - \sum_{j_s=1}^{N-n+1} c_{s,j_s} g_{s,j_s} = 0,
$$

which implies that there exists a vector in the reduced Gröbner basis of the syzygy of $\mathcal P$ with respect to the POT extension of the Lex order with $y^{(0)} \succ_L ... \succ_L y^{(N)} \succ_L u^{(0)} \succ_L ... \succ_L u^{(N)}$ whose last entry is a polynomial in $\mathbb{R}[u_{e,N}]$. □

Theorem 2 provides necessary and sufficient conditions for the existence of a simultaneous inverse $\Phi_N(\cdot, \cdot)$ of the observability maps $\Psi_{k,N}(\cdot, \cdot)$, $k = 1, ..., s$, together with computational tools to compute a closed-form expression of such an inverse. stence of a simultaneous
ps $\Psi_{k,N}(\cdot, \cdot)$, $k = 1, ...,$
npute a closed-form express
The next example illustrar
Example 2: Let $s = 2$ and
 $A_1 = \begin{bmatrix} 1 & -1 \\ 1 & 1 \end{bmatrix}, B_1 =$

The next example illustrates the application of such a technique.

$$
A_1 = \begin{bmatrix} 1 & -1 \\ 1 & -1 \end{bmatrix}, \quad B_1 = \begin{bmatrix} 0 \\ 1 \end{bmatrix}, \quad C_1 = \begin{bmatrix} 1 & 1 \end{bmatrix}, \quad D_1 = 1,
$$

\n
$$
A_2 = \begin{bmatrix} 1 & 2 \\ 1 & -1 \end{bmatrix}, \quad B_2 = \begin{bmatrix} 1 \\ 0 \end{bmatrix}, \quad C_2 = \begin{bmatrix} 1 & 0 \end{bmatrix}, \quad D_2 = 1.
$$

\nBy letting $N = 4$, and computing the sets $\mathcal{C}_{k,4}$, $k = 1, 2$, as in (9),
\none obtains the following polynomials of $\mathbb{R}[\mathbf{y}_{e,4}, \mathbf{u}_{e,4}]$:
\n
$$
\begin{cases} g_{1,1} = \mathbf{y}^{(4)} + 2\mathbf{u}^{(2)} - \mathbf{u}^{(3)} - \mathbf{u}^{(4)}, \end{cases}
$$

one obtains the following polynomials of ℝ[$y_{e, 4}$, $u_{e, 4}$]:

$$
\mathcal{E}_{1,4}\begin{cases} g_{1,1} = y^{(4)} + 2u^{(2)} - u^{(3)} - u^{(4)}, \\ g_{1,2} = y^{(3)} + 2u^{(1)} - u^{(2)} - u^{(3)}, \\ g_{1,3} = y^{(2)} + 2u^{(0)} - u^{(1)} - u^{(2)}, \\ g_{2,1} = 3y^{(2)} - y^{(4)} - 2u^{(2)} + u^{(3)} + u^{(4)}, \\ g_{2,2} = 3y^{(1)} - y^{(3)} - 2u^{(1)} + u^{(2)} + u^{(3)}, \\ g_{2,3} = 9y^{(0)} - y^{(4)} - 6u^{(0)} + 3u^{(1)} + u^{(2)} + u^{(3)} + u^{(4)} \end{cases}
$$

.

On the other hand, by computing **Θ** as in ([12b](#page-3-0)), one obtains

$$
\Theta = \begin{bmatrix} -\frac{1}{2}y^{(0)} + \frac{1}{4}y^{(1)} + \frac{1}{4}u^{(0)} - \frac{1}{4}u^{(1)} \\ y^{(0)} - \frac{3}{4}y^{(1)} - \frac{1}{4}u^{(0)} + \frac{3}{4}u^{(1)} \end{bmatrix}.
$$

Finally, by computing the reduced Gröbner basis of the syzygy of $[W_4 \quad \Theta]$, one of its elements is:

$$
\begin{bmatrix}\n-2y^{(0)} + y^{(1)} + u^{(0)} - u^{(1)} \\
0 \\
6y^{(0)} - 3y^{(1)} - 3u^{(0)} + 3u^{(1)} \\
4y^{(0)} - 3y^{(1)} - u^{(0)} + 3u^{(1)} \\
0 \\
-12y^{(0)} + 9y^{(1)} + 3u^{(0)} - 9u^{(1)} \\
2y^{(0)} - y^{(1)} - u^{(0)} + u^{(1)} \\
0 \\
0 \\
-4y^{(0)} + 3y^{(1)} + u^{(0)} - 3u^{(1)} \\
0 \\
24u^{(0)} - 12u^{(1)} - 12u^{(2)}\n\end{bmatrix}.
$$

Therefore, since the last entry of such a vector is in ℝ[*u_{e,4}*], by Theorem 2, there exists a simultaneous inverse $\Phi_4(\cdot, \cdot)$ of the observability maps $\Psi_{k,4}(\cdot, \cdot)$, $k = 1, 2$, that holds for 'almost all' Theorem 2, there exists a simultaneous inverse $\Phi_4(\cdot, \cdot)$ of the Therefore, since the last entry of such a vector is in $\mathbb{R}[u_{e,4}]$, by
Theorem 2, there exists a simultaneous inverse $\Phi_4(\cdot, \cdot)$ of the
observability maps $\Psi_{k,4}(\cdot, \cdot)$, $k = 1, 2$, that holds for 'almost all'
 $u_{e,$

$$
x_1 = -\frac{(u^{(0)} - u^{(1)} - 2y^{(0)} + y^{(1)})(2u^{(2)} - u^{(3)} - u^{(4)} - 3y^{(2)} + y^{(4)})}{12(-2u^{(0)} + u^{(1)} + u^{(2)})}
$$

$$
x_2 = \frac{(u^{(0)} - 3u^{(1)} - 4y^{(0)} + 3y^{(1)})(2u^{(2)} - u^{(3)} - u^{(4)} - 3y^{(2)} + y^{(4)})}{12(-2u^{(0)} + u^{(1)} + u^{(2)})},
$$
and can be used for all times $t \in \mathbb{R}_{\geq 0}$ such that
$$
u_{e,4}(t) \notin \mathbb{V}(\omega),
$$

$$
u_{e,4}(t) \notin \mathbb{V}(\omega),\tag{14}
$$

where $\omega = 2u^{(0)} - u^{(1)} - u^{(2)}$. It is worth noticing that if *u*_{*e*, 4}(*t*) ∉ **W**(*ω*), (14)

where $\omega = 2u^{(0)} - u^{(1)} - u^{(2)}$. It is worth noticing that if
 $u_{e,4}(t) \in V(\omega)$ for all times $t \in \mathbb{R}_{\geq 0}$, then system [\(7\)](#page-3-0) may be

unobservable without knowing which of the tuples (unobservable without knowing which of the tuples (A_1, B_1, C_1, D_1) and (A_2, B_2, C_2, D_2) is governing the dynamics of the system. Indeed, letting

$$
u(t) = \left(\frac{c_0}{3} - \frac{c_1}{3}\right) \exp(-2t) + \left(\frac{2c_0}{3} + \frac{c_1}{3}\right) \exp(t),\tag{15}
$$

Indeed, letting
 $u(t) = \left(\frac{c_0}{3} - \frac{c_1}{3}\right) \exp(-2t) + \left(\frac{2c_0}{3} + \frac{c_1}{3}\right) \exp(t)$, (15)

one has that $u_{e,4}(t) \in V(\omega)$ for all $t \in \mathbb{R}$ and for all $c_1, c_2 \in \mathbb{R}$.

Hence, letting *u* be as in (15), if the dynamics Hence, letting u be as in (15), if the dynamics of system ([7](#page-3-0)) are governed by (A_1, B_1, C_1, D_1) and the initial condition is

$$
\mathbf{x}_{1,0} = \begin{bmatrix} -\frac{3c_0}{4} - \frac{c_1}{4} \\ \frac{c_1}{4} - \frac{c_0}{4} \end{bmatrix},
$$

then the state and output response of system ([7](#page-3-0)) are given by

$$
\mathbf{x}_{1}(t) = \begin{bmatrix} -\frac{2c_{0} + c_{1}}{3} \exp(t) + \frac{c_{1} - c_{0}}{12} \exp(-2t) \\ \frac{c_{1} - c_{0}}{4} \exp(-2t) \\ y_{1}(t) = 0, \end{bmatrix},
$$

respectively. On the other hand, letting *u* be the input given in (15), if the dynamics of system [\(7\)](#page-3-0) are governed by (A_2, B_2, C_2, D_2) and the initial condition of the system is

,

$$
\boldsymbol{x}_{2,0} = \begin{bmatrix} -c_0 \\ -\frac{c_1}{2} \end{bmatrix},
$$

then the state and output response of system [\(7\)](#page-3-0) are given by

$$
\mathbf{x}_{2}(t) = \begin{bmatrix} -\frac{2c_{0} + c_{1}}{3} \exp(t) + \frac{c_{1} - c_{0}}{3} \exp(-2t) \\ -\frac{2c_{0} + c_{1}}{6} \exp(t) + \frac{c_{0} - c_{1}}{3} \exp(-2t) \\ y_{2}(t) = 0, \end{bmatrix},
$$

respectively. Hence, since $y_1(t) = y_2$ $\left\{\n \begin{aligned}\n &\frac{c_0 - c_1}{3} \exp(-2t)\n \end{aligned}\n \right\},\n \text{(}t) \text{ for all } t \in \mathbb{R}_{\geq 0}, \text{ but } x_1(t) \in \mathbb{R}_{\geq 0}, \text{ it is not possible to on measurement of } t \text{ and } y \text{ by letting } t \in \mathbb{R}_{\geq 0} \text{ and } y \text{ is the value of } t \in \mathbb{R}_{\geq 0} \text{ and } y \text{ are obtain } t \text{ and } y \text{ is the value of } t \in \mathbb{R}_{\geq 0} \text{ and } y \text{ is the value of$ $\left[-\frac{2c_0 + c_1}{6} \exp(t) + \frac{c_0 - c_1}{3} \exp(-2t)\right]'$
 $y_2(t) = 0,$

respectively. Hence, since $y_1(t) = y_2(t)$ for all $t \in \mathbb{R}_{\geq 0}$, but $x_1(t)$

need not be equal to $x_2(t)$ for all $t \in \mathbb{R}_{\geq 0}$, it is not possible to

r reconstruct the state of system ([7](#page-3-0)) from measurements of *u* and *y* without knowing which one among the tuples (A_1, B_1, C_1, D_1) and (A_2, B_2, C_2, D_2) is governing its dynamics.

Finally, consider the following definition.

Definition 2: A polynomial $p \in \mathbb{R}[y_{e,N}, u_{e,N}]$ is an *embedding* without knowing which one am

(*A*₂, *B*₂, *C*₂, *D*₂) is governing its dy

Finally, consider the following
 Definition 2: A polynomial i

of system ([7](#page-3-0)) if, $\forall x \in \mathbb{R}^n$, $\forall u_{e,N}$ (
 p($\Psi_{k,N}(x, u_e)$) , $\forall u_{e,N} \in \mathbb{R}^{N+1}, \forall k \in \{1, ..., s\},\$

$$
p(\mathbf{\Psi}_{k,N}(\mathbf{x}, \mathbf{u}_{e,N}), \mathbf{u}_{e,N}) = 0.
$$

Note that Definition 2 generalises Definition 1 to deal with systems whose dynamical matrices are not known. Indeed, by the construction given above, it results that if $p \in \mathbb{R}[y_{e,N}, u_{e,N}]$ is an embedding of system ([7](#page-3-0)), then

$$
p(\mathbf{y}_{e,N}(t), \mathbf{u}_{e,N}(t)) = 0, \quad \forall t \in \mathbb{R}_{\geq 0},
$$

independently of which $(A_k, B_k, C_k, D_k) \in S$ is governing the dynamics of the system. The following proposition is a straightforward corollary of Theorem 1.

Proposition 1: Let the polynomials $g_{k,N,1},...,g_{k,N,\ell}$ be defined as in ([9](#page-3-0)), $k = 1, ..., s$. The set of all embeddings of system [\(7\)](#page-3-0) is

$$
\mathcal{T}_N = \bigcap_{k=1}^s \langle g_{k,N,1}, \dots, g_{k,N,\ell} \rangle. \tag{16}
$$

The interest in the set \mathcal{T}_N defined in (16) relies on the fact that it allows to construct several simultaneous inverses of the observability maps $\Psi_{k,N}(\cdot, \cdot)$, $k = 1, ..., s$. Indeed, by the same reasoning used to prove Corollary 1, letting the assumptions of Theorem 2 hold, letting $\Phi_N(\cdot, \cdot)$ be defined as in [\(13](#page-3-0)), and letting $\{\lambda_1, ..., \lambda_{\kappa}\}\subset \mathbb{R}[y_{e,N}, u_{e,N}]$ be the reduced Gröbner basis of the ideal \mathcal{T}_N given in (16), the function $(y_{e,N}, u_{e,N}) \mapsto \Phi_N(y_{e,N}, u_{e,N}) + \sum_{j=1}^{N} c_j(y_{e,N}, u_{e,N}) \lambda_j(y_{e,N}, u_{e,N}),$ Theorem 2 hold, letting $\Phi_N(\cdot, \cdot)$ be defined as $\{\lambda_1, ..., \lambda_x\} \subset \mathbb{R}[y_{e,N}, u_{e,N}]$ be the reduced Grödeal \mathcal{T}_N given in (16),
ideal \mathcal{T}_N given in (16),
 $(y_{e,N}, u_{e,N}) \mapsto \Phi_N(y_{e,N}, u_{e,N}) + \sum_{j=1}^x c_j(y_{e,N}, u_{e,N})$
where c_1 $n[y_{e,N}, u_{e,N}]$, is a simultaneous inverse of the maps $\Psi_{k,N}(\cdot, \cdot)$, $k = 1, ..., s$, that holds for 'almost all' \mathbf{u}_{e} _N.

5 The case of autonomous systems

In this section, the analysis carried out in Section [4](#page-3-0) is specialised to the case of the following autonomous systems:

$$
\dot{x}(t) = A \, x(t), \quad y(t) = C \, x(t). \tag{17}
$$

Assume that the matrices *A* and *C* are not known, but that it is known that pair (A, C) belongs to the set

$$
\hat{\hat{\mathcal{S}}} = \{ (A_1, C_1), ..., (A_s, C_s) \},
$$

Assume that the matrices **A** and **C** are not known, but that it is

known that pair (A, C) belongs to the set
 $\hat{\delta} = \{(A_1, C_1), ..., (A_s, C_s)\}$,

where $s \in \mathbb{Z}_{\geq 2}$. The results stated in Theorem 2 still hold for

system system (17), and the main objective of this section is to show such systems. Toward this end, consider the following example in which

the results of Theorem 2 are applied to a systems being autonomous. results of Theorem 2 are applied to a
 Example 3: Let *s* = 3 and consider the matrices
 $A_1 = \begin{bmatrix} 1 & -1 \\ 0 & 1 \end{bmatrix}, C_1 = \begin{bmatrix} 1 & 0 \end{bmatrix},$

$$
A_1 = \begin{bmatrix} 1 & -1 \\ 0 & 1 \end{bmatrix}, \quad C_1 = \begin{bmatrix} 1 & 0 \end{bmatrix}, \tag{18a}
$$

$$
A_2 = \begin{bmatrix} 0 & 1 \\ 2 & 1 \end{bmatrix}, \quad C_2 = \begin{bmatrix} 0 & 1 \end{bmatrix}
$$
 (18b)

$$
A_3 = \begin{bmatrix} 2 & 1 \\ -1 & 1 \end{bmatrix}, \quad C_3 = \begin{bmatrix} 1 & 1 \end{bmatrix}. \tag{18c}
$$

 $A_2 = \begin{bmatrix} 2 & 1 \\ 2 & 1 \end{bmatrix}, \quad C_2 = [0 \quad 1]$ (18b)
 $A_3 = \begin{bmatrix} 2 & 1 \\ -1 & 1 \end{bmatrix}, \quad C_3 = [1 \quad 1].$ (18c)

By letting *N* = 5, and computing the sets $\mathcal{C}_{k, 5}, k = 1, 2, 3$, as in [\(9\)](#page-3-0),

one obtains the following polynomials of

$$
\mathcal{G}_{1,5} \begin{cases} g_{1,1} = y^{(3)} - 2y^{(4)} + y^{(5)}, \\ g_{1,2} = y^{(2)} - 3y^{(4)} + 2y^{(5)}, \\ g_{1,3} = y^{(1)} - 4y^{(4)} + 3y^{(5)}, \\ g_{1,4} = y^{(0)} - 5y^{(4)} + 4y^{(5)}, \\ g_{2,5} \begin{cases} g_{2,1} = 2y^{(3)} + y^{(4)} - y^{(5)}, \\ g_{2,2} = 4y^{(2)} - 3y^{(4)} + y^{(5)}, \\ g_{2,3} = 8y^{(1)} + 5y^{(4)} - 3y^{(5)}, \\ g_{2,4} = 16y^{(0)} - 11y^{(4)} + 5y^{(5)}, \\ g_{3,5} \end{cases} \\ \mathcal{G}_{3,5} \begin{cases} g_{3,1} = 3y^{(3)} - 3y^{(4)} + y^{(5)}, \\ g_{3,3} = 9y^{(2)} - 2y^{(4)} + y^{(5)}, \\ g_{3,4} = 9y^{(0)} - y^{(4)} + 2y^{(5)}, \end{cases}
$$

Hence, by considering that

$$
\begin{bmatrix} g_{1,1} \\ g_{2,1} \\ g_{2,2} \\ g_{2,3} \\ g_{2,4} \end{bmatrix} = \begin{bmatrix} 0 & 0 & 0 & 1 & -2 & 1 \\ 0 & 0 & 1 & 0 & -3 & 2 \\ 0 & 0 & 0 & 2 & 1 & -1 \\ 0 & 0 & 4 & 0 & -3 & 1 \\ 0 & 8 & 0 & 0 & 5 & -3 \\ 16 & 0 & 0 & 0 & -11 & 5 \end{bmatrix} \begin{bmatrix} y^{(0)} \\ y^{(2)} \\ y^{(3)} \\ y^{(4)} \\ y^{(5)} \\ y^{(6)} \\ y^{(7)} \\ y^{(8)} \\ g_{3,1} \\ g_{3,2} \end{bmatrix}
$$

$$
\begin{bmatrix} g_{1,3} \\ g_{1,4} \\ g_{3,1} \\ g_{3,2} \\ g_{3,3} \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & 0 & -4 & 3 \\ 1 & 0 & 0 & 0 & -5 & 4 \\ 0 & 0 & 0 & 3 & -3 & 1 \\ 0 & 0 & 0 & 3 & -3 & 1 \\ 0 & 0 & 0 & 0 & -3 & 2 \\ 0 & 9 & 0 & 0 & -3 & 2 \\ 9 & 0 & 0 & 0 & -1 & 1 \end{bmatrix} \begin{bmatrix} y^{(0)} \\ y^{(1)} \\ y^{(2)} \\ y^{(3)} \\ y^{(4)} \\ y^{(5)} \end{bmatrix},
$$

and that the matrices on the right-hand side of the expressions above have full rank, by the same reasoning used in the proof of Theorem 1, one has that

$$
\langle g_{1,1}, g_{1,2}, g_{2,1}, g_{2,2}, g_{2,3}, g_{2,4} \rangle = \langle y^{(0)}, y^{(1)}, \dots, y^{(5)} \rangle, \langle g_{1,3}, g_{1,4}, g_{3,1}, g_{3,2}, g_{3,3}, g_{3,4} \rangle = \langle y^{(0)}, y^{(1)}, \dots, y^{(5)} \rangle.
$$

Therefore, it results that, letting W_5 be defined as in [\(12a](#page-3-0)),

$$
Img(W_5) = \langle y^{(0)}e_1^4, \ldots, y^{(5)}e_1^4, y^{(0)}e_2^4, \ldots, y^{(5)}e_4^4 \rangle,
$$

i.e., there exists a constant matrix $K \in \mathbb{R}^{24 \times 24}$ such that

$$
\mathbf{W}_5\mathbf{K} = \mathbf{I}_4 \otimes \mathbf{y}_{e,5}^{\top}.
$$

Hence, since the polynomial vector **Θ** has linear entries in *y*_{*e*, 1}, i.e.
there exists a constant matrix $K \in \mathbb{R}^{24 \times 24}$ such that
 $W_5 K = I_4 \otimes y_{e.5}^T$.
Hence, since the polynomial vector **Θ** has linear entries there exists a matrix $\Upsilon \in \mathbb{R}^{4 \times 2}$ such that

$$
\Theta = \Upsilon \, y_{e,1},
$$

by Lemma 2.2 of [\[31](#page-10-0)] and Lemma 4 of [[32\]](#page-10-0), there exists a constant vector in the syzygy of $[W_5 \quad \Theta]$ (the explicit expression of such a vector is omitted for brevity). Thus, by Theorem 2, there exists a simultaneous inverse $\Phi_5(\cdot)$ of the observability maps $\Psi_{k,5}(\cdot, \cdot)$, $k = 1, 2, 3$, that, by using [\(13](#page-3-0)), is given by

$$
\mathbf{\Phi}_{s}(\mathbf{y}_{e, s}) = \begin{bmatrix} -\frac{11}{14} & \frac{7}{2} & -\frac{5}{7} & -\frac{31}{14} & \frac{3}{2} & \frac{2}{7} \\ \frac{73}{14} & \frac{35}{4} & \frac{1}{28} & \frac{51}{7} & \frac{19}{4} & \frac{27}{28} \end{bmatrix} \mathbf{y}_{e, s}.
$$

In view of the construction illustrated in Example 3, the following theorem specialises the results given in Theorem 2 to the case of autonomous systems.

Theorem 3: Let $s \in \mathbb{Z}_{\ge 2}$. If $N \ge n s - 1$, then, for 'almost all'
Theorem 3: Let $s \in \mathbb{Z}_{\ge 2}$. If $N \ge n s - 1$, then, for 'almost all'
Theorem 3: Let $s \in \mathbb{Z}_{\ge 2}$. If $N \ge n s - 1$, then, for 'almost all'
. $A_1, ..., A_s \in \mathbb{R}^{n \times n}$ and $C_1, ..., C_s \in \mathbb{R}^{1 \times n}$, there exists a linear simultaneous inverse of the observability maps $x \mapsto O_{k,N} x$, $k = 1, \ldots, s$, i.e., there exists $L_N \in \mathbb{R}^{n \times (N+1)}$ such that

$$
L_N O_{k,N} x = x, \quad \forall x \in \mathbb{R}^n, k = 1, ..., s. \tag{19}
$$

Proof: First, note that, for 'almost all' $A_1, ..., A_s \in \mathbb{R}^{n \times n}$ and $C_1, ..., C_s \in \mathbb{R}^{1 \times n}$, pairs (A_k, C_k) are observable [[27\]](#page-10-0). Let $\mathcal{G}_{k,N}$ be defined as in ([9](#page-3-0)), $k = 1, ..., s$: in view of Remark 1, such sets are composed by exactly $N - n + 1$ linear polynomials for 'almost all' matrices $A_1, ..., A_s \in \mathbb{R}^{n \times n}$ and $C_1, ..., C_s \in \mathbb{R}^{1 \times n}$. By using the construction made in Example 3, there exists a matrix $H_k \in \mathbb{R}^{(N+1)\times(N+1)}$ such that

$$
\begin{bmatrix} g_{1,(k-2)n+1} & \cdots & g_{1,(k-1)n} & \cdots & g_{k,N,N-n+1} \end{bmatrix}^T = \mathbf{H}_k \mathbf{y}_{e,N},
$$

for $k = 2, ..., s$. Furthermore, such a matrix has full rank for 'almost all' $A_1, ..., A_s \in \mathbb{R}^{n \times n}$ and $C_1, ..., C_s \in \mathbb{R}^{1 \times n}$. Thus, by the same reasoning used in Example 3, it results that

$$
\text{Img}(W_N) = \langle y^{(0)} e_1^{n(s-1)}, \dots, y^{(N)} e_1^{n(s-1)}, y^{(0)} e_{n(s-1)}^{n(s-1)}, \dots, y^{(N)} e_{n(s-1)}^{n(s-1)} \rangle.
$$

Therefore, by Lemma 2.2 of [\[31](#page-10-0)–[33\]](#page-10-0), there exists a constant vector in the syzygy of $[W_N \quad \Theta]$; therefore, by ([13\)](#page-3-0), there exists a linear inverse of the observability maps $x \mapsto O_{k,N} x$, $k = 1, ..., s$, for 'almost all' matrices $A_1, ..., A_s \in \mathbb{R}^{n \times n}$ and $C_1, ..., C_s \in \mathbb{R}^{1 \times n}$.

Theorem 3 provides a 'generic' result on the number of timederivatives of the output that have to be taken into account in order to allow one to jointly invert the observability maps $x \mapsto O_{k} x$, $k = 1, \ldots, s$. Finally, the following proposition characterises the number of time-derivatives of the output to be taken into account to guarantee the existence of a linear embedding for system ([17\)](#page-5-0).

Proposition 2: Let the set $\hat{\mathcal{S}}$ be given and let the ideal \mathcal{T}_N of estimate of $\mathbf{x}(t)$. As a matter of fact, system (*k* = 1, ..., *s*. Finally, the following proposition characterises the number of time-derivatives of the output to be taken into account to guarantee the existence of a linear embedding for system (17).
Proposition 2: $p \in \mathcal{T}_N$.

Proof: Define
$$
A_e := \text{diag}(A_1, ..., A_s) \in \mathbb{R}^{(ns) \times (ns)}
$$
, $C_e := [C_1 \dots C_s] \in \mathbb{R}^{1 \times (ns)}$, and consider the extended system

$$
\boldsymbol{\zeta} = A_e \boldsymbol{\zeta}, \qquad \boldsymbol{\varpi} = C_e \boldsymbol{\zeta}. \tag{20}
$$

It can be easily derived that if p is an embedding of system (20) , then it is also an embedding of system [\(17](#page-5-0)). Thus, the result follows directly from Theorem 1. \square

The following example illustrates the application of Proposition 1.

Example 4: Consider again the matrices given in (18). By *Example 4:* Consider again the matrices given in (18). By letting *N* = 6, computing the sets $\mathcal{G}_{k,6}$, $k = 1, 2, 3$, as in [\(9\)](#page-3-0), and computing the ideal \mathcal{T}_N as in (16), one obtains the polynomial $p = 6y^{(0)} - 15y$ computing the ideal \mathcal{T}_N as in ([16\)](#page-5-0), one obtains the polynomial

$$
p = 6y^{(0)} - 15y^{(1)} + 8y^{(2)} + 9y^{(3)} - 13y^{(4)} + 6y^{(5)} - y^{(6)}
$$

is in the ideal \mathcal{T}_N and hence it is a (linear) embedding of system ([17\)](#page-5-0).

6 State observer design

In this section, it is shown how the tools given in Sections [4](#page-3-0) and [5](#page-5-0) can be used to design a state observer for systems ([7](#page-3-0)) and [\(17](#page-5-0)). In particular, in Section 6.1, it is shown how the methods proposed in Section [5](#page-5-0) can be used to design a state observer for system ([17\)](#page-5-0) without knowing which of the pairs $(A_1, C_1), ..., (A_s, C_s)$ is governing its dynamics, whereas in Section [6.2](#page-7-0), it is shown how the techniques given in Section [4](#page-3-0) can be used to design a state observer for system ([7](#page-3-0)), without knowing which of the tuples $(A_1, B_1, C_1, D_1), \ldots, (A_s, B_s, C_s, D_s)$ is governing its dynamics. Assumption 1 is supposed to hold throughout this section.

6.1 Design of state observers for multiple autonomous systems

Assume that $N \geq n s - 1$, let *L* be such that (19) holds, and let

$$
p = \alpha_0 y_0 + \dots + \alpha_N y_N + y_{N+1} = \boldsymbol{\alpha}^{\top} \, \boldsymbol{y}_{e,N} + y_{N+1},
$$

 $p = \alpha_0 y_0 + \dots + \alpha_N y_N + y_{N+1} = \boldsymbol{\alpha}^\top \boldsymbol{y}_{e,N} + y_{N+1}$,
be a linear embedding of system ([17\)](#page-5-0) in \mathcal{T}_{N+1} (note that, in view
of the results established in Section [5,](#page-5-0) such that *L* and $\boldsymbol{\alpha}$ exist for 'almost all' $A_1, ..., A_s \in \mathbb{R}^{n \times n}$ and $C_1, ..., C_s \in \mathbb{R}^{1 \times n}$). Since p is a linear embedding of system ([17\)](#page-5-0), independent of the pairs $(A_1, C_1), ..., (A_s, C_s)$ which is governing its dynamics, the timederivatives of the output *y* satisfy

$$
\dot{\mathbf{y}}_{e,N}(t) = \mathbf{V} \, \mathbf{y}_{e,N}(t),\tag{21}
$$

where

$$
V = \begin{bmatrix} \mathbf{0}_{N,1} & I_N \\ -\boldsymbol{\alpha}^\top & \end{bmatrix} . \tag{22}
$$

Thus, a state observer for system ([17\)](#page-5-0) that is independent of the pairs $(A_1, C_1), ..., (A_s, C_s)$ which is governing its dynamics can be 'generically' given by

$$
\dot{\hat{y}}_{e,N}(t) = (V - T Y) \hat{y}_{e,N}(t) + T y(t),
$$
\n(23a)
\n
$$
\hat{x}(t) = L \hat{y}_{e,N}(t),
$$
\n(23b)
\nwhere $Y = [1 \quad \mathbf{0}_{1,N}], T$ is such that $V - T Y$ has eigenvalues with

$$
\hat{\mathbf{x}}(t) = \mathbf{L} \,\hat{\mathbf{y}}_{e,N}(t),\tag{23b}
$$

negative real part, $\hat{y}_{e,N}(t)$ is an estimate of $y_{e,N}(t)$ and $\hat{x}(t)$ is an $\sum_{i=1}^{8}$ estimate of $x(t)$. As a matter of fact, system (23) is a classical Luenberger observer [[1\]](#page-10-0) for system (21) and hence the estimation error $\hat{y}_{e,N}(t) - y_{e,N}(t)$ converges exponentially to 0. Therefore, since $\mathbf{x}(t) = L \mathbf{y}_{e,N}(t)$, it results that also the estimation error $\hat{x}(t) - x(t)$ converges exponentially to 0, i.e., system (23) is an exponential state observer for system [\(17](#page-5-0)). System (23) is referred to as a simultaneous state observer for multiple autonomous systems. The next example illustrates the application of this observer.

Example 5: Consider the mechanical system depicted in Fig. [1,](#page-7-0) which is constituted by two bodies having mass m and two springs having stiffness k . Let the output y be the position of the first body. *Example 5:* Consider the mechanical system depicted in Fig. 1, which is constituted by two bodies having mass m and two springs having stiffness k. Let the output *y* be the position of the first body. Letting the masses *Example 5:* Consider the mechanical system which is constituted by two bodies having mas having stiffness k. Let the output *y* be the positi Letting the masses be unitary and assuming whether the stiffness is $k = 1$ or

Letting the masses be unitary and assuming that it is not knwon

Fig. 1 *A mechanical system with two masses and two springs*

Fig. 2 *Time behaviour of the state of the mechanical system, time history of the state of the state observer (23), and estimation error* $\tilde{x} = x - \hat{x}$ **Fig. 2** Time behaviour of the state of the state of the state observable when $k = 1$

Fig. 3 *Time behaviour of the state of the mechanical system, time history of the state of the observer (23), and estimation error* $\tilde{x} = x - \hat{x}$ *obtained* 40 ^U
Fig. 3 *Time behavior*
*of the state of the obs
when* $k = 2$

$$
A_1 = \begin{bmatrix} 0 & 1 & 0 & 0 \\ -2 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 1 & 0 & -1 & 0 \end{bmatrix}, A_2 = \begin{bmatrix} 0 & 1 & 0 & 0 \\ -4 & 0 & 2 & 0 \\ 0 & 0 & 0 & 1 \\ 2 & 0 & -2 & 0 \end{bmatrix},
$$

$$
C_1 = \begin{bmatrix} 1 & 0 & 0 & 0 \end{bmatrix}, C_2 = \begin{bmatrix} 1 & 0 & 0 & 0 \end{bmatrix}.
$$

Using the results established in Section [5](#page-5-0) (see Examples 3 and 4 for the explicit steps that have to be carried out in order to compute the matrix L and the vector α), one obtains

$$
L = \begin{bmatrix} 1 & 0 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 & 0 & 0 \\ \frac{4}{3} & 0 & -\frac{7}{6} & 0 & -\frac{7}{6} & 0 & -\frac{1}{6} & 0 \\ 0 & \frac{4}{3} & 0 & -\frac{7}{6} & 0 & -\frac{7}{6} & 0 & -\frac{1}{6} \end{bmatrix},
$$

$$
\alpha^{T} = \begin{bmatrix} -4 & 0 & -18 & 0 & -23 & 0 & -9 & 0 \end{bmatrix}.
$$

Hence, letting *V* be defined as in [\(22](#page-6-0)), one has that the matrix (see equation below) is such that $V - T Y$ has eigenvalues with negative $\begin{bmatrix} 0 & \frac{4}{3} & 0 & -\frac{7}{6} & 0 & -\frac{7}{6} & 0 & -\frac{1}{6} \end{bmatrix}$
 $\alpha^T = [-4 \ 0 \ -18 \ 0 \ -23 \ 0 \ -9 \ 0].$

Hence, letting *V* be defined as in (22), one has that the matrix (see

equation below) is such that *V* − *T Y* has eige real part. Thus, the simultaneous state observer (23) provides an Hence, letting V be defined as in (22), one has that the matrix (see equation below) is such that $V - TY$ has eigenvalues with negative real part. Thus, the simultaneous state observer (23) provides an exponentially conver

Numerical simulations have been carried out to test such an observer. Fig. 2 depicts the behaviour of the mechanical system, the time history of the state of the observer (23) and the estimation exponentially converging estimate of the state of
system depicted in Fig. 1 independently of whether Numerical simulations have been carried out
observer. Fig. 2 depicts the behaviour of the mec
the time history of the st $\hat{\mathbf{y}}_{e,7}(0) = \mathbf{0}_{7,1}$. On the other hand, Fig. 3 depicts the time behaviour of the mechanical system, the time history of the state of the observer observer. Fig. 2 depicts the behaviour of the mechanical system,
the time history of the state of the observer (23) and the estimation
error obtained when $k = 1$, $\mathbf{x}(0) = \begin{bmatrix} 1 & 0 & -1 & 0 \end{bmatrix}^T$, $\hat{\mathbf{y}}_{e.7}(0) = \mathbf{$

$$
\mathbf{x}(0) = \begin{bmatrix} 1 & 0 & -1 & 0 \end{bmatrix}^\top, \hat{\mathbf{y}}_{e,7}(0) = \mathbf{0}_{7,1}.
$$

In order to compare the proposed observer with the ones existing in the literature, the simultaneous observation scheme given in Section III of [[13\]](#page-10-0) has been implemented for the mechanical system depicted in Fig. 1 by designing the observer gains as for the observer (23). Figs. [4](#page-8-0) and [5](#page-8-0) depict the results of numerical simulations of such an observer in which the system has been initialised at the same initial condition of the ones reported in existing in the literature, the simultaneous observation
given in Section III of [13] has been implemented
mechanical system depicted in Fig. 1 by designing the
gains as for the observer (23). Figs. 4 and 5 depict the
num

As shown by such figures, although the transient behaviour of the observer given in Section III of [\[13](#page-10-0)] is similar to the one of the observer (23) and both are capable of estimating the state of the mechanical system without knowing whether the stiffness of the Figs. 2 and 3 and the observer has been initialised at $\mathbf{0}_{12,1}$.
As shown by such figures, although the transient behaviour of
the observer given in Section III of [\[13](#page-10-0)] is similar to the one of the
observer (23) and state dimension; as a matter of fact, it has 12 states whereas the observer (23) has eight states. Nonetheless, the computations to be performed off-line to design the proposed observer are slightly more complex than the ones required to design the observer given in Section III of [\[13](#page-10-0)]. In fact, while the latter can be designed by determining two stabilising gains for two linear plants, the former **r** (0 1 0 0 0 0
 $A = \begin{bmatrix} 0 & 1 & 0 & 0 \\ -2 & 0 & 1 & 0 \\ 0 & 1 & 0 & 0 \\ 1 & 1 & -1 & 0 \\ 0 & 1 & 0 & 1 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\$ (see Examples 3 and 4 for the explicit steps that have to be carried out) and, second, the computation of the stabilising gain *T*.

6.2 Design of state observers for multiple systems with inputs

When dealing with systems with inputs, the design strategy proposed in Section [6.1](#page-6-0) cannot be directly applied due to the fact that there need not exist a linear embedding for system ([7](#page-3-0)) that holds independently of which of the tuples $(A_1, B_1, C_1, D_1), \ldots, (A_s, B_s, C_s, D_s)$ is governing its dynamics. However, it is still possible to design a state observer for such a system by using the results given in [[34\]](#page-10-0).

Let $N \in \mathbb{Z}_{\geq 0}$ be fixed so that there exists a simultaneous inverse $\Phi_N(\cdot, \cdot)$ of the observability maps $\Psi_{k,N}(\cdot, \cdot)$,

,

$$
T = [16.64 \quad 121.3 \quad 470.1 \quad 830.3 \quad -508.1 \quad -4601 \quad -1830 \quad 21,922]^\top
$$

Fig. 4 *Time history of the state of the observer given in Section III of [\[13](#page-10-0)] and estimation error* $\tilde{x} = x - \hat{x}$ *with* $k = 1$

Fig. 5 *Time history of the state of the observer given in Section III of [\[13](#page-10-0)] and estimation error* $\tilde{x} = x - \hat{x}$ *with* $k = 2$

Fig. 6 *An electric circuit with two switches*

 $k = 1, \ldots, s$, that holds for 'almost all' $u_{e,N}$ (see Theorem 2 for the conditions ensuring that such an assumption hold). Hence, **Fig. 6** An electric circuit with two switches
 $k = 1, ..., s$, that holds for 'almost all' $u_{e,N}$ (see Theorem 2 for the

conditions ensuring that such an assumption hold). Hence,

assuming that, for all times $t \in \mathbb{R}_{\geq$ d^{N+1} $\left|\frac{d}{dt^{N+1}}u(t)\right| \leq \Delta$ for some $\Delta \in \mathbb{R}_{\geq 0}$ and that the corresponding state response of system [\(7\)](#page-3-0) is such that $|x(t)| \leq \Lambda$ for some $\Lambda \in \mathbb{R}_{\geq 0}$, an observer for system ([7](#page-3-0)) that is independent of which of the tuples $(A_1, B_1, C_1, D_1), \ldots, (A_s, B_s, C_s, D_s)$ is governing its dynamics is 'generically' given by

$$
\dot{\hat{y}}^{(0)}(t) = \hat{y}^{(1)}(t) + \frac{\kappa_1}{\varepsilon} (y(t) - \hat{y}^{(0)}(t)),
$$
\n(24a)

$$
\dot{\hat{y}}^{(1)}(t) = \hat{y}^{(2)}(t) + \frac{\kappa_2}{\varepsilon^2} (y(t) - \hat{y}^{(0)}(t)),
$$
\n(24b)

$$
\dot{y}^{(N)}(t) = \frac{\kappa_{N+1}}{\varepsilon^{N+1}} (y(t) - \dot{y}^{(0)}(t)),
$$
\n(24c)

$$
\dot{\hat{u}}^{(0)}(t) = \hat{u}^{(1)}(t) + \frac{\kappa_1}{\varepsilon}(u(t) - \hat{u}^{(0)}(t)),
$$
\n(24d)

$$
\dot{\hat{u}}^{(1)}(t) = \hat{u}^{(2)}(t) + \frac{\kappa_2}{\varepsilon^2} (u(t) - \hat{u}^{(0)}(t)),
$$
\n
$$
\vdots
$$
\n(24e)

$$
\dot{\hat{u}}^{(N)}(t) = \frac{\kappa_{N+1}}{\varepsilon^{N+1}} (u - \hat{u}^{(0)}),
$$
\n(24f)

$$
\hat{\mathbf{x}}(t) = \mathbf{\Phi}_N(\hat{\mathbf{y}}_{e,N}(t), \hat{\mathbf{u}}_{e,N}(t))
$$
\n(24g)

where $\hat{y}_{e,N}(t) = \begin{bmatrix} y^{(0)}(t) & \dots & y^{(N)}(t) \end{bmatrix}^\top$ is the estimate of $y_{e,N}(t)$, $\hat{u}_{e,N}(t) = \begin{bmatrix} u^{(0)}(t) & \dots & u^{(N)}(t) \end{bmatrix}^\top$ is the estimate of $u_{e,N}(t)$, the coefficients $\kappa_1, \ldots, \kappa_{N+1}$ are such that the polynomial $\zeta^{N+1} + \kappa_1 \zeta^N + \dots + \kappa_N \sigma + \kappa_{N+1}$ is Hurwitz, *ε* is a sufficiently small positive real parameter, and $\hat{\mathbf{x}}(t)$ is an estimate of $\mathbf{x}(t)$. As a matter of fact, by [\[34](#page-10-0)], if $|d^{N+1}/dt^{N+1}$, $|u(t)| \le \Delta$ and $|x(t)| \le \Delta$ (thus implying by Lemma 1 that there exists $\bar{\Lambda}$ such that $\frac{a}{b}$ $\left|\frac{d^{N+1}}{dt^{N+1}}\right| u(t) \leq \bar{\Lambda}$), then the estimation errors $y_{e,N}(t) - \hat{y}_{e,N}(t)$ and $u_{e,N}(t) - \hat{u}_{e,N}(t)$ can be made arbitrarily small in an arbitrarily small amount of time by letting the parameter ε be sufficiently small. Therefore, in view of the absolute continuity of the function $\Phi_N(\cdot, \cdot)$ in its domain, system (24) is able of 'practically' estimate (i.e. with arbitrarily small estimation error and convergence time) the state of system [\(7\)](#page-3-0), without requiring the knowledge of which of the tuples $(A_1, B_1, C_1, D_1), \ldots, (A_s, B_s, C_s, D_s)$ is governing its dynamics. Such a system is referred to as a simultaneous state observer for multiple systems with inputs.

The next example illustrates the application of such an observer.

Example 6: Consider the electric circuit depicted in Fig. 6, which is composed by a capacitor with capacitance c , an inductor with inductance 1 two resistors with resistance r , and two switches. Assume that the position of the two switches is not known and let the output *y* be the voltage across the resistor and let the input *u* be the current delivered by the current source.

Letting the values of the parameters be unitary, the dynamics of the circuit are modelled by system ([7](#page-3-0)) with $s = 3$,

$$
A_1 = \begin{bmatrix} 0 & 1 \\ -\frac{1}{2} & 0 \end{bmatrix}, \quad B_1 = \begin{bmatrix} 0 \\ 1 \end{bmatrix}, \quad C_1 = \begin{bmatrix} 0 & 1 \end{bmatrix},
$$

$$
A_2 = \begin{bmatrix} 0 & 1 \\ -\frac{1}{2} & -1 \end{bmatrix}, \quad B_2 = \begin{bmatrix} 0 \\ 1 \end{bmatrix}, \quad C_2 = \begin{bmatrix} 0 & 1 \end{bmatrix},
$$

$$
A_3 = \begin{bmatrix} 0 & 1 \\ -\frac{1}{2} & -\frac{1}{2} \end{bmatrix}, \quad B_3 = \begin{bmatrix} 0 \\ 1 \end{bmatrix}, \quad C_3 = \begin{bmatrix} 0 & 1 \end{bmatrix},
$$

and $D_1 = D_2 = D_3 = 0$, corresponding to the possible configurations of the switches. Letting $N = 6$ and using the results $A_3 = \begin{bmatrix} 0 & 1 \\ -\frac{1}{2} & -\frac{1}{2} \end{bmatrix}$, $B_3 = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$, $C_3 = \begin{bmatrix} 0 & 1 \end{bmatrix}$,
and $D_1 = D_2 = D_3 = 0$, corresponding to the possible
configurations of the switches. Letting $N = 6$ and using the results
establi established in Section [4](#page-3-0) (see Example 2 for the explicit steps that have to be carried out), one obtains a simultaneous inverse $\Phi_6(\cdot, \cdot)$ of the observability maps $\Psi_{k, 6}(\cdot, \cdot)$, $k = 1, 2, 3$, (whose explicit expression is omitted for compactness), which can be used for all input functions $u(t)$ such that $u^{(2)}(t)$ does not vanish (or vanishes at isolated time instants). Hence, such a function can be used to design the simultaneous state observer (24).

Fig. 7 *Time behaviour of the state of the electrical circuit and estimation* $error \tilde{x} = x - \hat{x}$ *obtained when no switch is closed*

Fig. 8 *Time behaviour of the state of the electrical circuit and estimation* $error \tilde{x} = x - \hat{x}$ *obtained when one switch is closed*

Fig. 9 *Time behaviour of the state of the electrical circuit and estimation* $error \tilde{x} = x - \hat{x}$ *obtained when two switches are closed*

Numerical simulations have been carried out to test such an observer. Fig. 7 depicts the time behaviour of the state of the electrical circuit and the estimation error obtained by using the error $\tilde{x} = x - \hat{x}$ obtained when two switches

Numerical simulations have been

observer. Fig. 7 depicts the time be

electrical circuit and the estimation of

state observer (24) with $\varepsilon = 10^{-3}$, $\hat{y}_{e,6}$
 $(\kappa_3 =$, *y* ^ *error* $\tilde{x} = x - \hat{x}$ *obtained when two switches are closed*

Numerical simulations have been carried out to test such an

observer. Fig. 7 depicts the time behaviour of the state of the

electrical circuit and the esti $\mathbf{x}(0) = \begin{bmatrix} 1 \\ 0 \end{bmatrix}^\mathsf{T}, u(t) = 0.01 t^2.$

Fig. 8 depicts the behaviour of the electrical circuit system and the estimation error obtained by using the state observer (24) with the same parameters as above, when one switch is closed, $x(0) = [1 \ 0]^T$ and $u(t) = 0.01 t^2$.

Finally, Fig. 9 depicts the behaviour of the electrical circuit system and the estimation error obtained by using the observer (24) with the same parameters as above, when two switches are closed, $\mathbf{x}(0) = \begin{bmatrix} 1 \\ 0 \end{bmatrix}^T$ and $u(t) = 0.01 t^2$.

As shown by Figs. 7–9, the simultaneous state observer (24) is capable of practically reconstructing the state of the electrical system without requiring the knowledge of the position of the switches.

Note that the results given in [\[13](#page-10-0)] cannot be directly applied to design an observer for the considered electrical circuit since [[13\]](#page-10-0) does not deal with the case of three systems with inputs. Furthermore, also the approach given in [[16\]](#page-10-0) cannot be directly used since Assumption A2 of [\[16](#page-10-0)] does not hold for the considered system.

7 Conclusions

In this paper, algebraic geometry tools have been proposed to characterise the simultaneous observability of a set of linear systems and to design a simultaneous state observer. In order to pursue this objective, first an algebraic technique has been proposed to compute the set of all the embeddings of a single linear system and it has been used to find a parametrisation of all the linear inverses of its observability map. Such a parametrisation has been used to provide necessary and sufficient conditions for the existence of a simultaneous inverse, holding for almost all inputs, of the observability maps of multiple linear systems and to provide an algebraic geometry technique capable of computing such an inverse. The results given for SISO linear system have been then specialised to the case of autonomous systems, for which much stronger results hold. In particular, a 'generic' result has been given on the number of time-derivatives of the output that have to be taken into account in order to allow one to jointly invert *k* observability maps. Finally, it has been shown how such techniques can be directly used to design simultaneous observers for multiple linear systems by coupling them with 'practical' high-gain observers. The theoretical results have been corroborated and illustrated by several examples reported all throughout the paper.

The main advantage of the tools given in this paper with respect to others given in the literature $[9, 11-13]$ $[9, 11-13]$ $[9, 11-13]$ $[9, 11-13]$ $[9, 11-13]$ $[9, 11-13]$ is that they provide an exact certificate for the simultaneous observability of a set of linear system or for the lack thereof. Furthermore, they allow one to compute directly a closed-form expression for the simultaneous inverse of the observability maps of the systems in the set, which can be readily used to design simultaneous observers by interfacing it with high-gain observers.

It is worth pointing out that although, for simplicity, the results given in Section [6.2](#page-7-0) have been illustrated assuming that the input is C^k for some sufficiently large $k \in \mathbb{Z}_{\geq 0}$, the proposed technique can erse of the observability maps of the systems in the set, which
the readily used to design simultaneous observers by interfacing
vith high-gain observers.
It is worth pointing out that although, for simplicity, the result be employed also if the input is discontinuous, provided that there is a minimum dwell time between two consecutive discontinuities (see [[24\]](#page-10-0) for further details).

As shown in Example 5, the proposed simultaneous state observer for multiple autonomous systems has state dimension smaller than others available in the literature, although the computations that have to be carried out off-line to design the observer may be slightly more complex. Furthermore, as shown in Example 6, the given simultaneous state observer for multiple systems with inputs can be used in some cases in which other design procedures cannot be used, although it requires more strict hypotheses on the input, which has to be piecewise C^k for some $\sum_{n=1}^{\infty}$ observer may be slightly more complex. Furthermore, as shown in Example 6, the given simultaneous state observer for multiple systems with inputs can be used in some cases in which other design procedures cannot be used, two consecutive discontinuities.

It is worth mentioning that the proposed algebraic technique for a single system can be extended so to deal with unknown inputs. In fact, by [[35,](#page-10-0) [36\]](#page-10-0), the state *ξ* of system [\(1\)](#page-1-0) is observable with unknown inputs if and only if system ([1](#page-1-0)) is differentially flat. In such a case, by letting $\mathcal{I}_N := \langle \psi_{e,N} - O_N \xi - M_N \psi_{e,N} \rangle$ be the ideal generated by the relations given in ([3](#page-2-0)), an inverse of the observability map $\Psi_N(\xi, \nu_{e,N})$ that is independent of the input ν and

its time derivatives can be determined by computing the Gröbner basis of $\mathcal{K}_N := \mathcal{F}_N \cap \mathbb{R}[\xi, \psi_{e,N}]$ according to the Lex order with $\psi^{(0)}$ >_L ... >_L $\psi^{(N)}$ >_L ξ_1 >_L ... ξ_n . In particular, there exists a rational inverse of the observability map $\Psi_N(\xi, \nu_{e,N})$ that is independent of the input ν and its time derivatives if and only if there are polynomials $g_1, ..., g_n \in \mathcal{K}_N$ such that $LT(g_i) = \varphi(\psi_{e,N})x_i$, $i = 1, \ldots, n$ (see [24] for further details). Designing a common unknown input observer is therefore, easy with the method proposed here if such polynomials exist, as shown in the following example. 1,...,*n* (see [24] for further details). Design
mown input observer is therefore, easy we
posed here if such polynomials exist, as shown
mple.
Example 7: Consider system [\(7](#page-3-0)) with $s = 2$,
 $\begin{bmatrix} -\frac{1}{3} & \frac{2}{3} \end{bmatrix}$

$$
A_1 = \begin{bmatrix} -\frac{1}{3} & \frac{2}{3} \\ -\frac{5}{3} & \frac{1}{3} \end{bmatrix}, \quad B_1 = \begin{bmatrix} -3 \\ 3 \end{bmatrix}, \quad C_1 = \begin{bmatrix} \frac{1}{3} & \frac{1}{3} \end{bmatrix},
$$

$$
A_2 = \begin{bmatrix} \frac{7}{3} & \frac{10}{3} \\ -\frac{13}{3} & -\frac{7}{3} \end{bmatrix}, \quad B_2 = \begin{bmatrix} -1 \\ 1 \end{bmatrix}, \quad C_2 = \begin{bmatrix} \frac{1}{3} & \frac{1}{3} \end{bmatrix}.
$$

By defining the ideal $\mathcal{I}_{1,1} := \langle y_{e,1} - \mathbf{O}_{1,1} x - M_{1,1} u_{e,1} \rangle$ and computing the reduced Gröbner basis of $\mathcal{K}_{1,1} := \mathcal{F}_{1,1} \cap \mathbb{R}[x, y_{e,1}]$ according to the Lex order with $y^{(0)} >_L y^{(1)} >_L x_1 >_L x_2$, one obtains that $\mathcal{K}_{1,1} = \langle g_1, g_2 \rangle$ with

$$
g_1 = x_2 - 2y_0 - y_1,
$$

$$
g_2 = x_1 - y_0 + y_1,
$$

which implies that

$$
\boldsymbol{\theta}_1 = \begin{bmatrix} y_0 - y_1 \\ 2y_0 + y_1 \end{bmatrix}
$$

is an inverse of the observability map $\Psi_{1,1}(x, u_{e,1})$ that is independent of the input u and its time derivative. Furthermore, by computing the reduced Gröbner basis of the syzygy of $[W_3 \quad \Theta]$, one of its elements is $\begin{bmatrix} 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 \end{bmatrix}^T$, and hence θ_1 is a common inverse of the observability maps $\Psi_{1,3}(x, u_{e,3})$ and $\Psi_{2,3}(x, u_{e,3})$ that is independent of the input *u* and its time derivatives. Therefore, a common unknown input observer for the considered system can be designed by coupling the observer ([24a\)](#page-8-0)– ([24c\)](#page-8-0) with such an inverse.

On the other hand, if there do not exist polynomials $g_{k,1}, \ldots, g_{k,n} \in \langle y_{e,N} - O_{k,N} x - M_{k,N} u_{e,N} \rangle \cap \mathbb{R}[y_{e,N}, x]$ such that $LT(g_{k,i}) = \varphi(\psi_{e,N})x_i$, $i = 1, ..., n$, $k = 1, ..., s$, it is not easy to extend the techniques proposed in this paper; this would correspond to design unknown input observers for multiple systems in the case of unknown input detectability, as in [14–16].

8 References

- [1] O'Reilly, J.: '*Observers for linear systems*' (Academic Press, London, UK, 1983)
- [2] Sussmann, H.J.: 'Generic single-input observability of continuous-time polynomial systems'. IEEE Conf. Decision and Control including 17th Symp. Adaptive Processes, Fort Lauderdale, FL, USA, 1979, vol. 17, pp. 566–571
- [3] Aeyels, D.: 'Generic observability of differentiable systems', *SIAM J. Control Optim.*, 1981, **19**, pp. 595–603 [4] Gauthier, J.-P., Kupka, I.: '*Deterministic observation theory and applications*'
- (Cambridge University Press, Cambridge, UK, 2001)
- [5] Menini, L., Possieri, C., Tornambe, A.: 'Sinusoidal disturbance rejection in chaotic planar oscillators', *Int. J. Adapt. Control Signal Process.*, 2015, **29**, (12), pp. 1578–1590
- [6] Menini, L., Possieri, C., Tornambe, A.: 'A 'practical' observer for nonlinear systems'. 56th Conf. Decision Control, Melbourne, VIC, Australia, 2017, pp. 3015–3020
- [7] Luenberger, D.: 'Observers for multivariable systems', *IEEE Trans. Autom. Control*, 1966, **11**, (2), pp. 190–197
- [8] Kalman, R.E.: 'A new approach to linear filtering and prediction problems', *J. Basic Eng.*, 1960, **82**, (1), pp. 35–45
- [9] Yao, Y.X., Darouach, M., Schaefers, J.: 'Simultaneous observation of linear systems', *IEEE Trans. Autom. Control*, 1995, **40**, (4), pp. 696–699
- [10] Yao, Y.X., Radun, A.V.: 'Proportional integral observer design for linear systems with time delay', *IET Control Theory Applic.*, 2007, **1**, (4), pp. 887– 892
- [11] Fernandez-Anaya, G., Munoz-Futierrez, S., Sanchez-Guzman, R.*, et al.*: 'Simultaneous and robust observability using evolutionary strategies and substitutions'. 38th Conf. Decision and Control, Phoenix, AZ, USA, 1999, vol. 1, pp. 79–84
- [12] Kovacevic, R., Yao, Y., Zhang, Y.: 'Observer parameterization for simultaneous observation', *IEEE Trans. Autom. Control*, 1996, **41**, (2), pp. 255–259
- [13] Moreno, J.A.: 'Simultaneous observation of linear systems: a state-space interpretation', *IEEE Trans. Autom. Control*, 2005, **50**, (7), pp. 1021–1025
- [14] Trinh, H., Teh, P.S., Ha, Q.P.: 'A common disturbance decoupled observer for linear systems with unknown inputs', *Int. J. Autom. Control*, 2008, **2**, (2-3), pp. 286–297
- [15] Ng, J.Y., Tan, C.P., Ng, K.Y.*, et al.*: 'New results in common functional state estimation for two linear systems with unknown inputs', *Int. J. Control Autom. Syst.*, 2015, **13**, (6), pp. 1538–1543
- [16] Ng, J.Y., Tan, C.P., Trinh, H.*, et al.*: 'A common functional observer scheme for three systems with unknown inputs', *J. Franklin Inst.*, 2016, **353**, (10), pp. 2237–2257
- [17] Alessandri, A., Coletta, P.: 'Switching observers for continuous-time and discrete-time linear systems', *Am. Control Conf.*, 2001, **3**, pp. 2516–2521
- [18] Chen, W., Mehrdad, S.: 'Observer design for linear switched control systems', *Am. Control Conf.*, 2004, **6**, pp. 5796–5801
- [19] Pettersson, S.: 'Observer design for switched systems using multiple quadratic Lyapunov functions'. Mediterranian Conf. Control Automation Intelligence and Control, Limassol, Cyprus, 2005, pp. 262–267
- [20] Gómez-Gutiérrez, D., Čelikovskỳ, S., Ramírez-Treviño, A.*, et al.*: 'Sliding mode observer for switched linear systems'. Int. Conf. Automation and Science Engineering, Trieste, Italy, 2011, pp. 725–730
- [21] Cox, D.A., Little, J., O'Shea, D.: '*Ideals, varieties, and algorithms*' (Springer, New York, USA, 2015)
- [22] Cox, D.A., Little, J., O'Shea, D.: '*Using algebraic geometry*' (Springer, New York, USA, 2006)
- [23] Meyer, C.D.: '*Matrix analysis and applied linear algebra*' (SIAM, Philadelphia, PA, USA, 2000)
- [24] Menini, L., Possieri, C., Tornambe, A.: 'Switching signal estimator design for a class of elementary systems', *IEEE Trans. Autom. Control*, 2016, **61**, (5), pp. 1362–1367
- [25] Menini, L., Possieri, C., Tornambe, A.: 'Application of algebraic geometry techniques in permanent-magnet DC motor fault detection and identification', *Eur. J. Control*, 2015, **25**, pp. 39–50
- [26] Menini, L., Possieri, C., Tornambe, A.: 'On observer design for a class of continuous-time affine switched or switching systems'. 53rd Conf. Decision and Control, Los Angeles, CA, USA, 2014, pp. 6234–6239
- [27] Kailath, T.: '*Linear systems*' (Prentice-Hall, New Jersey, USA, 1980)
- [28] Menini, L., Possieri, C., Tornambe, A.: 'Observers for linear systems by the time-integrals and moving average of the output', *IEEE Trans. Autom. Control*, 2019. DOI: 10.1109/TAC.2019.2901827
- [29] Buchberger, B.: 'A theoretical basis for the reduction of polynomials to canonical forms', *ACM SIGSAM Bull.*, 1976, **10**, (3), pp. 19–29
- [30] Lazard, D.: 'Gröbner bases, Gaussian elimination and resolution of systems of algebraic equations'. Computer Algebra, London, UK, 1983, pp. 146–156
- [31] Menini, L., Tornambe, A.: 'On a Lyapunov equation for polynomial continuous-time systems', *Int. J. Control*, 2014, **87**, (2), pp. 393–403
- [32] Possieri, C., Tornambe, A.: 'On polynomial vector fields having a given affine variety as attractive and invariant set: application to robotics', *Int. J. Control*, 2015, **88**, (5), pp. 1001–1025
- [33] Possieri, C., Tornambe, A.: 'On f-invariant and attractive affine varieties for continuous-time polynomial systems: the case of robot motion planning'. 54th Conf. Decision and Control, 2014, pp. 3751–3756
- [34] Tornambe, A.: 'High-gain observers for non-linear systems', *Int. J. Syst. Sci.*, 1992, **23**, (9), pp. 1475–1489
- [35] Barbot, J.-P., Fliess, M., Floquet, T.: 'An algebraic framework for the design of nonlinear observers with unknown inputs'. 46th IEEE Conf. Decision and Control, New Orleans, LA, USA, 2007, pp. 384–389
- [36] Fliess, M., Join, C., Sira-Ramirez, H.: 'Non-linear estimation is easy', *Int. J. Modell. Identif. Control*, 2008, **4**, (1), pp. 12–27